

ECE/ENGRD 2100

Introduction to Circuits for ECE

Lecture 36

Transfer Function in the s-Domain
and Poles and Zeros

Announcements

- Recommended Reading:
 - Textbook Chapter 13 and Appendix E
- Upcoming due dates:
 - Homework 5 due by 11:59 pm on Monday April 29, 2019
 - Prelab 6 due by 11:59 pm on Monday April 29, 2019
 - Lab report 6 due by 11:59 pm on Friday May 3, 2019
 - Homework 6 due by 11:59 pm on Tuesday May 7, 2019

Laplace Transform Properties

Time Domain

Frequency Domain

$$K_1 f_1(t) + K_2 f_2(t)$$



$$K_1 F_1(s) + K_2 F_2(s)$$



$$\frac{df(t)}{dt}$$



$$sF(s) - f(0^-)$$



$$\frac{d^2 f(t)}{dt^2}$$



$$s^2 F(s) - sf(0^-) - f'(0^-)$$

$$\int_{-\infty}^t f(t) dt$$



$$\frac{F(s)}{s} + \frac{1}{s} \int_{-\infty}^{0^-} f(t) dt$$

$$f(t - t_0)u(t - t_0)$$



$$e^{-t_0 s} F(s)$$

$$e^{-\alpha t} f(t)$$



$$F(s + \alpha)$$

$$f(0^+)$$



$$\lim_{s \rightarrow \infty} sF(s)$$

$$f(\infty)$$

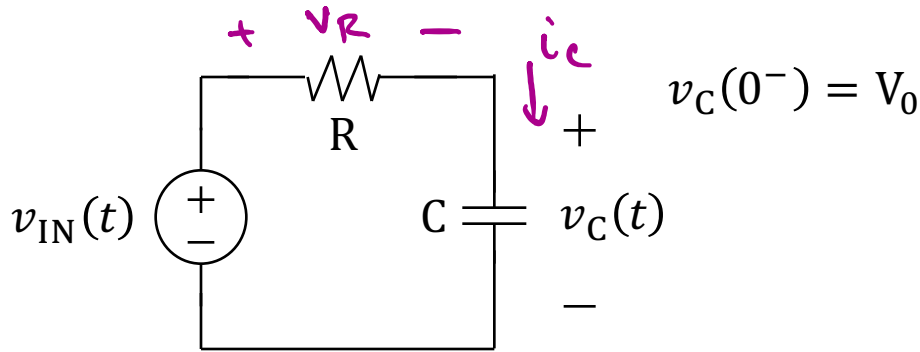


$$\lim_{s \rightarrow 0} sF(s)$$

Laplace Transform Pairs

Time Domain		Frequency Domain	
$\delta(t)$	\longleftrightarrow	1	\leftarrow
$u(t)$	\longleftrightarrow	$\frac{1}{s}$	\leftarrow
$tu(t)$	\longleftrightarrow	$\frac{1}{s^2}$	
$e^{-\alpha t}u(t)$	\longleftrightarrow	$\frac{1}{s + \alpha}$	
$\sin(\omega t) u(t)$	\longleftrightarrow	$\frac{\omega}{s^2 + \omega^2}$	
$\cos(\omega t) u(t)$	\longleftrightarrow	$\frac{s}{s^2 + \omega^2}$	
$te^{-\alpha t}u(t)$	\longleftrightarrow	$\frac{1}{(s + \alpha)^2}$	
$e^{-\alpha t} \sin(\omega t) u(t)$	\longleftrightarrow	$\frac{\omega}{(s + \alpha)^2 + \omega^2}$	
$e^{-\alpha t} \cos(\omega t) u(t)$	\longleftrightarrow	$\frac{s + \alpha}{(s + \alpha)^2 + \omega^2}$	

Example 2 – Step Response with Initial Condition



$$v_R + v_C = v_{IN}$$

$$RC \frac{dv_C}{dt} + v_C = V_I u(t)$$

↓ \mathcal{L}

$$v_{IN}(t) = V_I u(t)$$

$$RC(sV_C(s) - v_C(0^-)) + V_C(s) = \frac{V_I}{s}$$

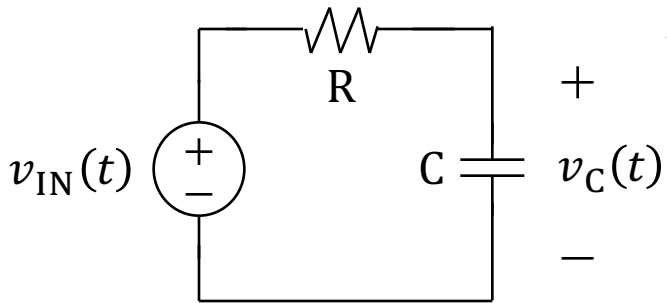
$$\Rightarrow (RCs + 1)V_C(s) = \frac{V_I}{s} + RCv_C(0^-)$$

$$V_C(s) = \frac{V_I/RC}{s(s + 1/RC)} + \frac{v_C(0^-)}{s + 1/RC}$$

←

$$V_C(s) = \frac{A_1}{s} + \frac{A_2}{s + 1/RC} + \frac{v_C(0^-)}{s + 1/RC}$$

Example 2 (Cont.)



$$V_C(s) = \frac{V_I/RC}{s(s + 1/RC)} + \frac{V_0}{s + 1/RC}$$

$$V_C(s) = \frac{A_1}{s} + \frac{A_2}{s + 1/RC} + \frac{V_0}{s + 1/RC}$$

$$v_{IN}(t) = V_I u(t)$$

$$\frac{A_1 s + A_1/RC + A_2 s}{s(s + 1/RC)} = \frac{V_I/RC}{s(s + 1/RC)}$$

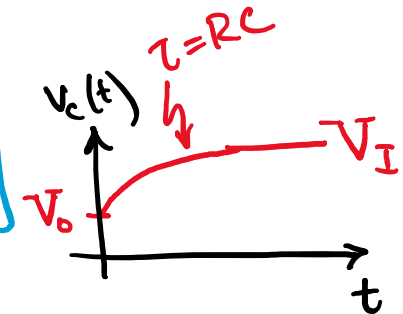
$$A_1 + A_2 = 0 \quad \& \quad A_1/RC = V_I/RC \Rightarrow \boxed{A_1 = V_I}$$

$$\Rightarrow A_2 = -A_1 \Rightarrow \boxed{A_2 = -V_I}$$

$$V_C(s) = \frac{V_I}{s} - \frac{V_I}{s + 1/RC} + \frac{V_0}{s + 1/RC}$$

$$\Leftrightarrow v_C(t) = V_I u(t) - V_I e^{-t/RC} u(t) + V_0 e^{-t/RC} u(t)$$

$$\Rightarrow \boxed{v_C(t) = V_I (1 - e^{-t/RC}) u(t) + V_0 e^{-t/RC} u(t)}$$



Partial Fractions – Faster Method

$$\frac{V_I/RC}{s(s + 1/RC)} \equiv \frac{A_1}{s} + \frac{A_2}{s + 1/RC}$$

To find A_1 , multiply by s
& set $s=0$

$$\Rightarrow \frac{V_I/RC \cdot s}{s(s+1/RC)} = \frac{A_1 \cdot s}{s} + \frac{A_2 \cdot s}{s+1/RC}$$

$$\Rightarrow \underline{A_1 = V_I}$$

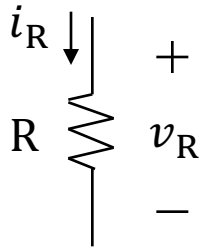
To find A_2 , multiply by $(s+1/RC)$
& set $s = -1/RC$

$$\Rightarrow \frac{V_I/RC}{s} = \frac{A_1(s+1/RC)}{s} + A_2$$

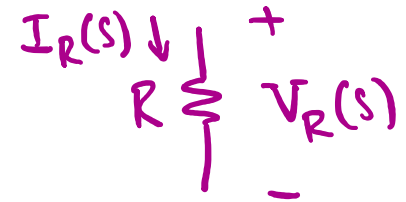
$$\Rightarrow \underline{A_2 = -V_I}$$

Laplace Transform of Components

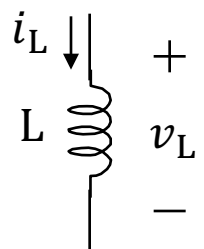
Resistor



$$v_R = Ri_R \iff \mathcal{L}\{v_R\} = \mathcal{L}\{Ri_R\} \Rightarrow \underline{V_R(s)} = R \underline{I_R(s)}$$



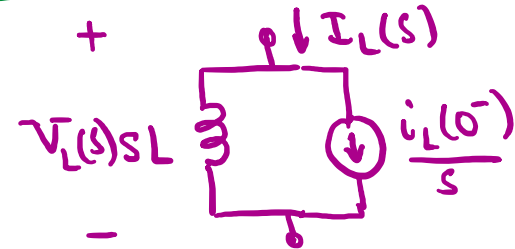
Inductor



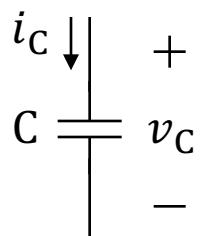
$$v_L = L \frac{di_L}{dt} \iff \mathcal{L}\{v_L\} = \mathcal{L}\left\{L \frac{di_L}{dt}\right\} \Rightarrow \underline{V_L(s)} = L(s \underline{I_L(s)} - i_L(0^-))$$

$$i_L(0^-) = I_0$$

$$\underline{I_L(s)} = \frac{\underline{V_L(s)}}{sL} + \frac{i_L(0^-)}{s}$$



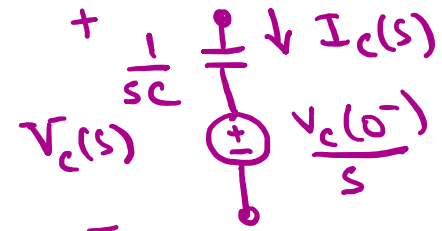
Capacitor



$$i_C = C \frac{dv_C}{dt} \iff \mathcal{L}\{i_C\} = \mathcal{L}\left\{C \frac{dv_C}{dt}\right\} \Rightarrow \underline{I_C(s)} = C(s \underline{V_C(s)} - v_C(0^-))$$

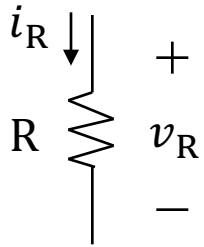
$$v_C(0^-) = V_0$$

$$\underline{V_C(s)} = \frac{\underline{I_C(s)}}{sC} + \frac{v_C(0^-)}{s}$$

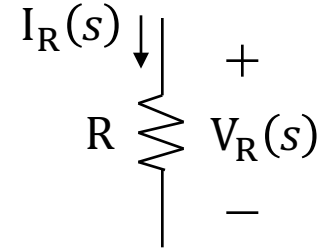


Laplace Transform of Components – Summary

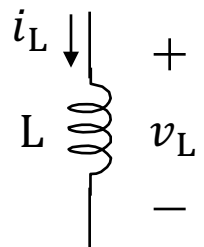
Resistor



$$v_R = Ri_R \iff V_R(s) = RI_R(s)$$

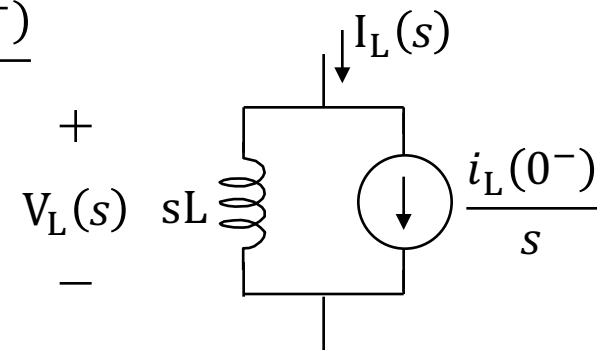


Inductor

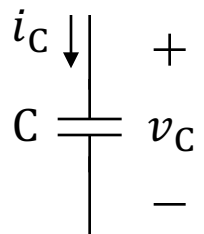


$$v_L = L \frac{di_L}{dt} \iff I_L(s) = \frac{V_L(s)}{sL} + \frac{i_L(0^-)}{s}$$

$$i_L(0^-) = I_0$$

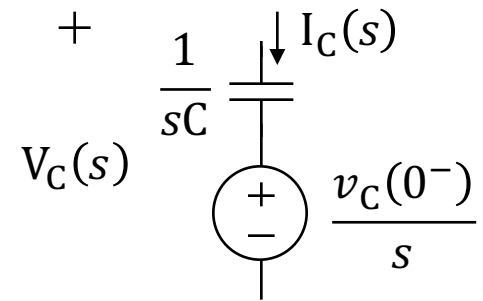


Capacitor



$$i_C = C \frac{dv_C}{dt} \iff V_C(s) = \frac{I_C(s)}{sC} + \frac{v_C(0^-)}{s}$$

$$v_C(0^-) = V_0$$



Laplace Transform of Circuit Laws

Kirchhoff's Voltage Law

$$\sum_{k=1}^N v_k(t) = 0 \quad \longleftrightarrow \quad \mathcal{L}\left\{\sum_{k=1}^N v_k(t)\right\} = 0 \quad \Rightarrow \quad \sum_{k=1}^N \mathcal{L}\{v_k(t)\} = 0$$
$$\Rightarrow \sum_{k=1}^N V_k(s) = 0$$

Kirchhoff's Current Law

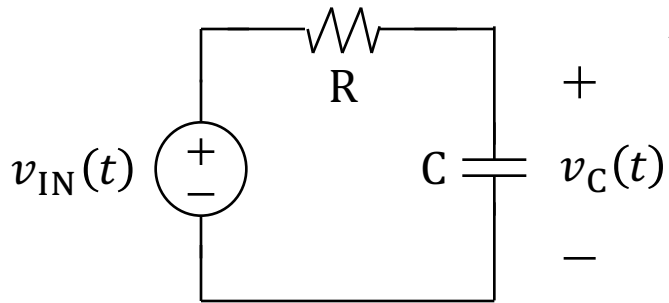
$$\sum_{k=1}^M i_k(t) = 0 \quad \longleftrightarrow \quad \mathcal{L}\left\{\sum_{k=1}^M i_k(t)\right\} = 0 \quad \Rightarrow \quad \sum_{k=1}^M \mathcal{L}\{i_k(t)\} = 0$$
$$\Rightarrow \sum_{k=1}^M I_k(s) = 0$$

Circuit Analysis using Laplace Transform

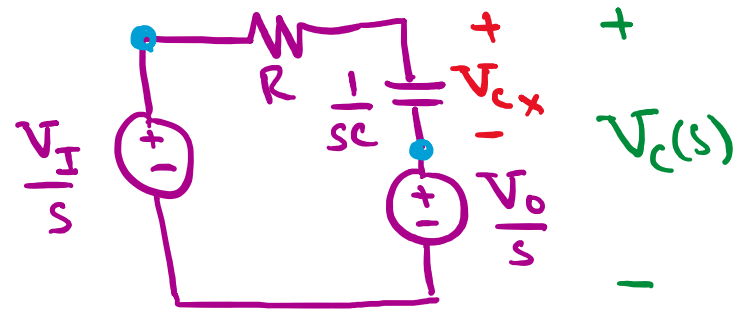
1. Create s -domain (frequency domain) model of the circuit
 - Replace sources by their Laplace transform

 - Replace circuit elements by their s -domain impedance models
2. Solve s -domain circuit for s -domain variables of interest
3. Convert s -domain variables of interest into time domain by taking inverse Laplace transform
 - Use partial fraction expansion to convert s -domain result into form with known inverse Laplace transform

Example 2 (Again)



$$v_C(0^-) = V_0$$



$$v_{IN}(t) = V_I u(t)$$

$$V_{C_x} = \frac{\frac{1}{sC}}{\frac{1}{sC} + R} \left(\frac{V_I}{s} - \frac{V_0}{s} \right) = \frac{1}{(1+sRC)s} (V_I - V_0)$$

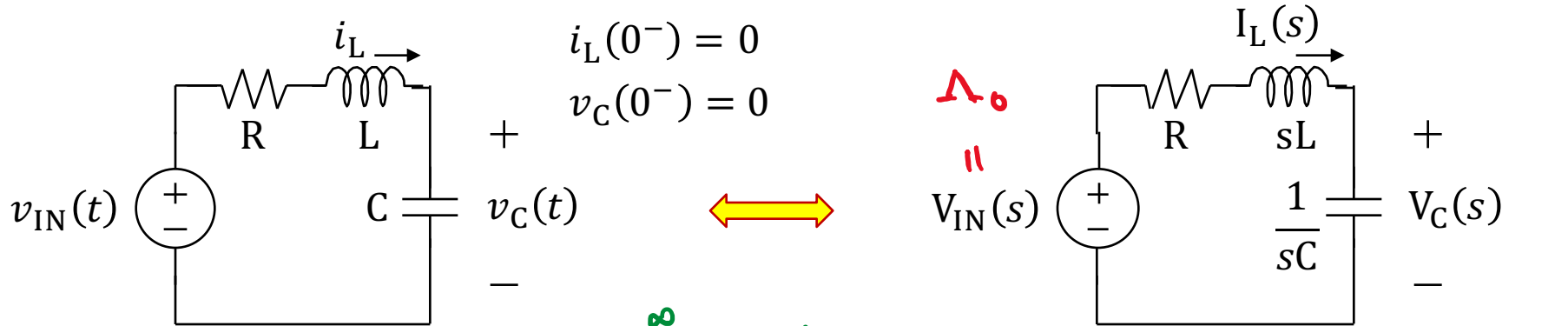
$$\frac{-\cancel{V_0} + \cancel{V_0} + sRCV_0}{s(1+sRC)}$$

$$V_C(s) = V_{C_x} + \frac{V_0}{s} = \frac{V_I}{s(1+sRC)} - \frac{V_0}{s(1+sRC)} + \frac{V_0}{s}$$

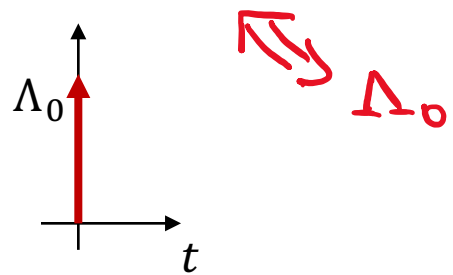
$$V_C(s) = \frac{V_I/RC}{s(s + 1/RC)} + \frac{V_0}{s + 1/RC}$$

$$\rightarrow V_C(s) = \frac{V_I/RC}{s(s + 1/RC)} + \frac{V_0}{s + 1/RC}$$

Example 3 – Impulse Response in s-Domain



$$v_{IN}(t) = \Lambda_0 \delta(t)$$



$$\mathcal{L}\{s(t)\} = \int_{-\infty}^{\infty} s(t) e^{-st} dt$$

$$= \int_{0^-}^{0^+} \delta(t) e^0 dt = 1$$

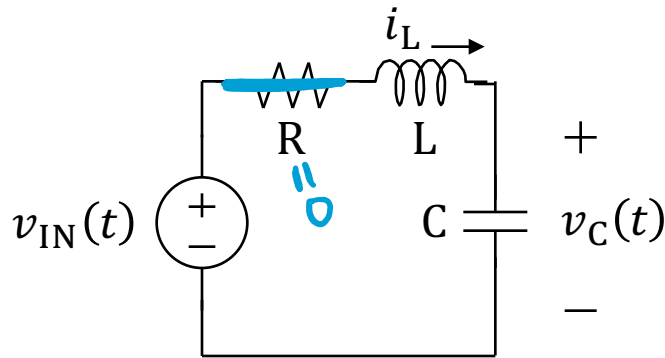
$$I_L(s) = \frac{\Lambda_0}{sL + R + 1/sC}$$

$$I_L(s) = \frac{s\Lambda_0 C}{s^2 LC + sRC + 1}$$

Example: $R = 0$ & define $\frac{1}{\sqrt{LC}} = \omega_0$

$$I_L(s) = \frac{s\Lambda_0 C}{s^2 LC + 1} = \frac{s\Lambda_0/L}{s^2 + 1/LC} = \frac{s\Lambda_0/L}{s^2 + \omega_0^2}$$

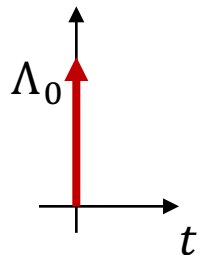
Example 3 (Cont.)



$$\cos(\omega t) u(t) \longleftrightarrow \frac{s}{s^2 + \omega^2}$$

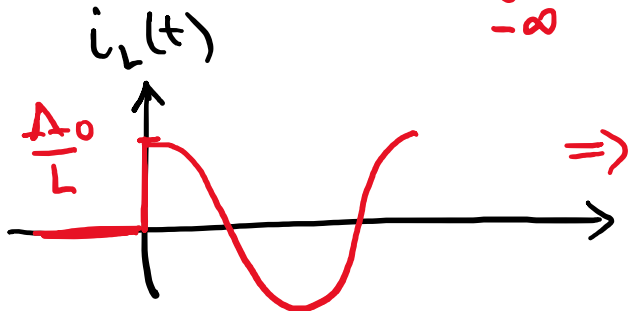
$$I_L(s) = \frac{s \Lambda_0 / L}{s^2 + \omega_0^2}$$

$$v_{IN}(t) = \Lambda_0 \delta(t)$$



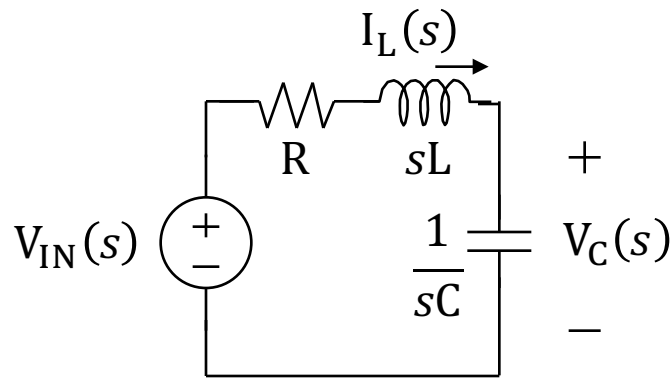
$$\Rightarrow i_L(t) = \frac{\Lambda_0}{L} \cos(\omega_0 t) u(t)$$

$$i_L(0^+) = \frac{1}{L} \int_{-\infty}^{0^+} v_L(t) dt = \frac{1}{L} \int_{0^-}^{0^+} \Lambda_0 \delta(t) dt = \frac{\Lambda_0}{L}$$



$$\Rightarrow \underline{\underline{\frac{\Lambda_0}{L} \cos(\omega_0 t) u(t)}}$$

Transfer Function in s-Domain



$$H(s) \equiv \frac{Y(s)}{X(s)}$$

output (any variable of interest)
with initial conditions = 0
input (drive)

$$H_{IL}(s) \equiv \frac{I_L(s)}{V_{IN}(s)} = \frac{\frac{s}{L}}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$$

$$H_{VC}(s) \equiv \frac{V_C(s)}{V_{IN}(s)}$$


Poles and Zeros of the Transfer Function

$$H(s) \equiv \frac{Y(s)}{X(s)} = \frac{N(s)}{D(s)}$$

Poles are the roots of the denominator polynomial – $D(s)$

Zeros are the roots of the numerator polynomial – $N(s)$

$$Y(s) = H(s)X(s) = \frac{N(s)}{D(s)}X(s)$$

Impulse Response: $X(s) = 1$  $Y(s) = H(s) = \frac{N(s)}{D(s)}$