

ECE 2300
Digital Logic & Computer Organization
Fall 2016

Combinational Logic Minimization



Cornell University

Lecture 3: 1

Example: Prime Number Detector (?)

- $F = 1$ if number xyz is prime
- Step 1: Lay out truth table
- Step 2: Derive canonical form
 - $F = \sum_{x,y,z}(1,2,3,5,7) = x'y'z + x'yz' + x'yz + xy'z + xyz$
 - $F = \prod_{x,y,z}(0,4,6) = (x+y+z)(x'+y+z)(x'+y'+z)$
- Step 3: Simplify expression
 - Algebraic simplification
 - Systematic minimization (next time)

xyz	F
000	0
001	1
010	1
011	1
100	0
101	1
110	0
111	1

Algebraic Simplification

- $F = \sum_{x,y,z}(1,2,3,5,7)$

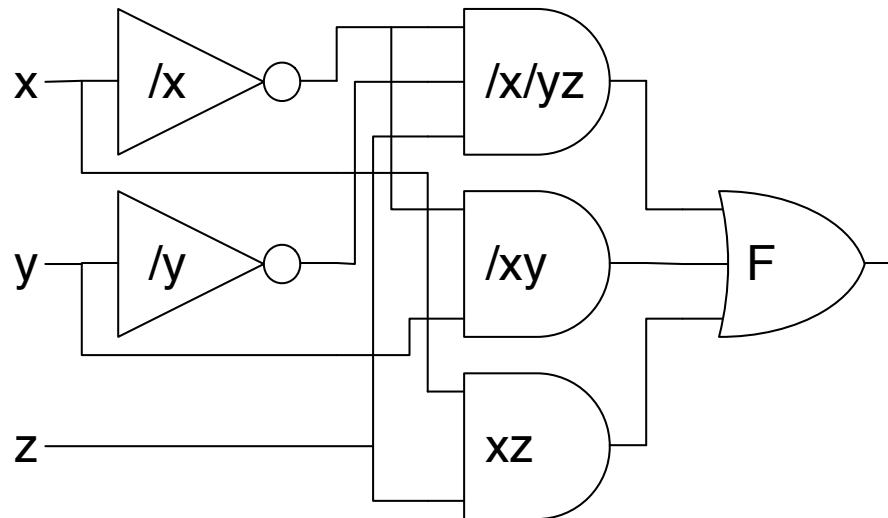
$$= x'y'z + x'yz' + x'yz + xy'z + xyz$$

$$= x'y'z + x'y + xy'z + xyz$$

[combining]

$$= x'y'z + x'y + xz$$

[combining]



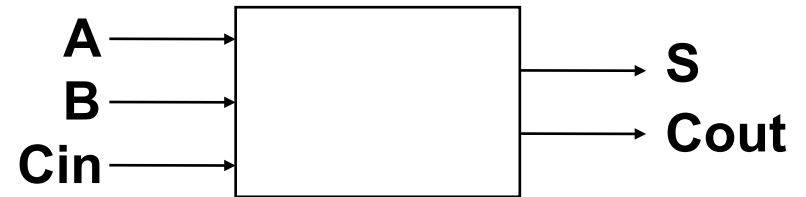
Combinational Logic

- **Outputs depend only on current inputs**
 - **Example: Detect if the input is a prime number**
- **In contrast, *sequential logic* has “memory” or “state”**
 - **Example: Detect if the last two inputs are prime**
 - **We’ll cover sequential logic later**

Algebraic Simplification Example

- **1-bit binary adder**

- inputs: A, B, Carry-in
- outputs: Sum, Carry-out



- **Truth Table → Canonical sum**

A	B	Cin	S	Cout
0	0	0	0	0
0	0	1	1	0
0	1	0	1	0
0	1	1	0	1
1	0	0	1	0
1	0	1	0	1
1	1	0	0	1
1	1	1	1	1

$$S = A' \cdot B' \cdot Cin + A' \cdot B \cdot Cin' + A \cdot B' \cdot Cin' + A \cdot B \cdot Cin$$

$$Cout = A' \cdot B \cdot Cin + A \cdot B' \cdot Cin + A \cdot B \cdot Cin' + A \cdot B \cdot Cin$$

Idempotency and Combining Theorems

- Idempotency: $X+X = X$
- Combining: $X \cdot Y + X \cdot Y' = X$

Algebraic Simplification Example

$$\begin{aligned} \text{Cout} &= A' \cdot B \cdot \text{Cin} + A \cdot B' \cdot \text{Cin} + A \cdot B \cdot \text{Cin}' + A \cdot B \cdot \text{Cin} \\ &= A' \cdot B \cdot \text{Cin} + A \cdot B' \cdot \text{Cin} + A \cdot B \cdot \text{Cin}' + A \cdot B \cdot \text{Cin} + A \cdot B \cdot \text{Cin} \quad (\text{idempotency}) \\ &= B \cdot \text{Cin} + A \cdot B' \cdot \text{Cin} + A \cdot B \cdot \text{Cin}' + A \cdot B \cdot \text{Cin} \quad (\text{combining}) \\ &= B \cdot \text{Cin} + A \cdot B' \cdot \text{Cin} + A \cdot B \cdot \text{Cin}' + A \cdot B \cdot \text{Cin} + A \cdot B \cdot \text{Cin} \quad (\text{idempotency}) \\ &= B \cdot \text{Cin} + A \cdot \text{Cin} + A \cdot B \cdot \text{Cin}' + A \cdot B \cdot \text{Cin} \quad (\text{combining}) \\ &= B \cdot \text{Cin} + A \cdot \text{Cin} + A \cdot B \quad (\text{combining}) \end{aligned}$$

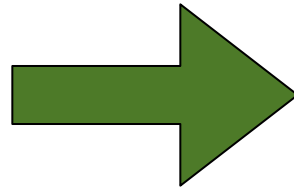
We can apply these theorems in a more systematic fashion using a *Karnaugh Map*

Karnaugh Map

- **Idea: Use combining and idempotency theorems *visually* to simplify canonical forms**
- **Multidimensional representation of a truth table**
- **Adjacent cells represent minterms (or maxterms) that differ by one variable**
 - **Cyclic encoding along each dimension**
- **At most two variables per dimension**

Karnaugh Map for Cout

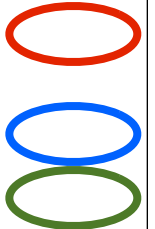
A	B	Cin	S	Cout
0	0	0	0	0
0	0	1	1	0
0	1	0	1	0
0	1	1	0	1
1	0	0	1	0
1	0	1	0	1
1	1	0	0	1
1	1	1	1	1



		AB			
		00	01	11	10
Cin	0	0	0	1	0
	1	0	1	1	1

Simplifying Cout Using a K Map

$$\begin{aligned}
 \text{Cout} &= A' \cdot B \cdot \text{Cin} + A \cdot B' \cdot \text{Cin} + A \cdot B \cdot \text{Cin}' + A \cdot B \cdot \text{Cin} \\
 &= A' \cdot B \cdot \text{Cin} + A \cdot B' \cdot \text{Cin} + A \cdot B \cdot \text{Cin}' + \color{blue}{A \cdot B \cdot \text{Cin}} + \color{blue}{A \cdot B \cdot \text{Cin}} && \text{(idempotency)} \\
 &= \color{blue}{B \cdot \text{Cin}} + A \cdot B' \cdot \text{Cin} + A \cdot B \cdot \text{Cin}' + A \cdot B \cdot \text{Cin} && \text{(combining)} \\
 &= B \cdot \text{Cin} + A \cdot B' \cdot \text{Cin} + A \cdot B \cdot \text{Cin}' + \color{blue}{A \cdot B \cdot \text{Cin}} + \color{blue}{A \cdot B \cdot \text{Cin}} && \text{(idempotency)} \\
 &= B \cdot \text{Cin} + \color{blue}{A \cdot \text{Cin}} + A \cdot B \cdot \text{Cin}' + A \cdot B \cdot \text{Cin} && \text{(combining)} \\
 &= B \cdot \text{Cin} + A \cdot \text{Cin} + \color{blue}{A \cdot B} && \text{(combining)}
 \end{aligned}$$



		AB			
		00	01	11	10
Cin	0	0	0	1	0
	1	0	1	1	1

Ovals result from applying combining theorem

Idempotency theorem allows ovals to overlap

Some K Map Definitions

- 1-cell: Minterm of a canonical sum
- Implicant: Set of adjacent 1-cells
 - Must be a power of 2
- Prime Implicant: Implicant that cannot be contained in a larger implicant
- Distinguished 1-cell: 1-cell covered by only one prime implicant (*essential prime implicant*)

		AB			
		00	01	11	10
Cin	0	0	0	1	0
	1	0	1	1	1

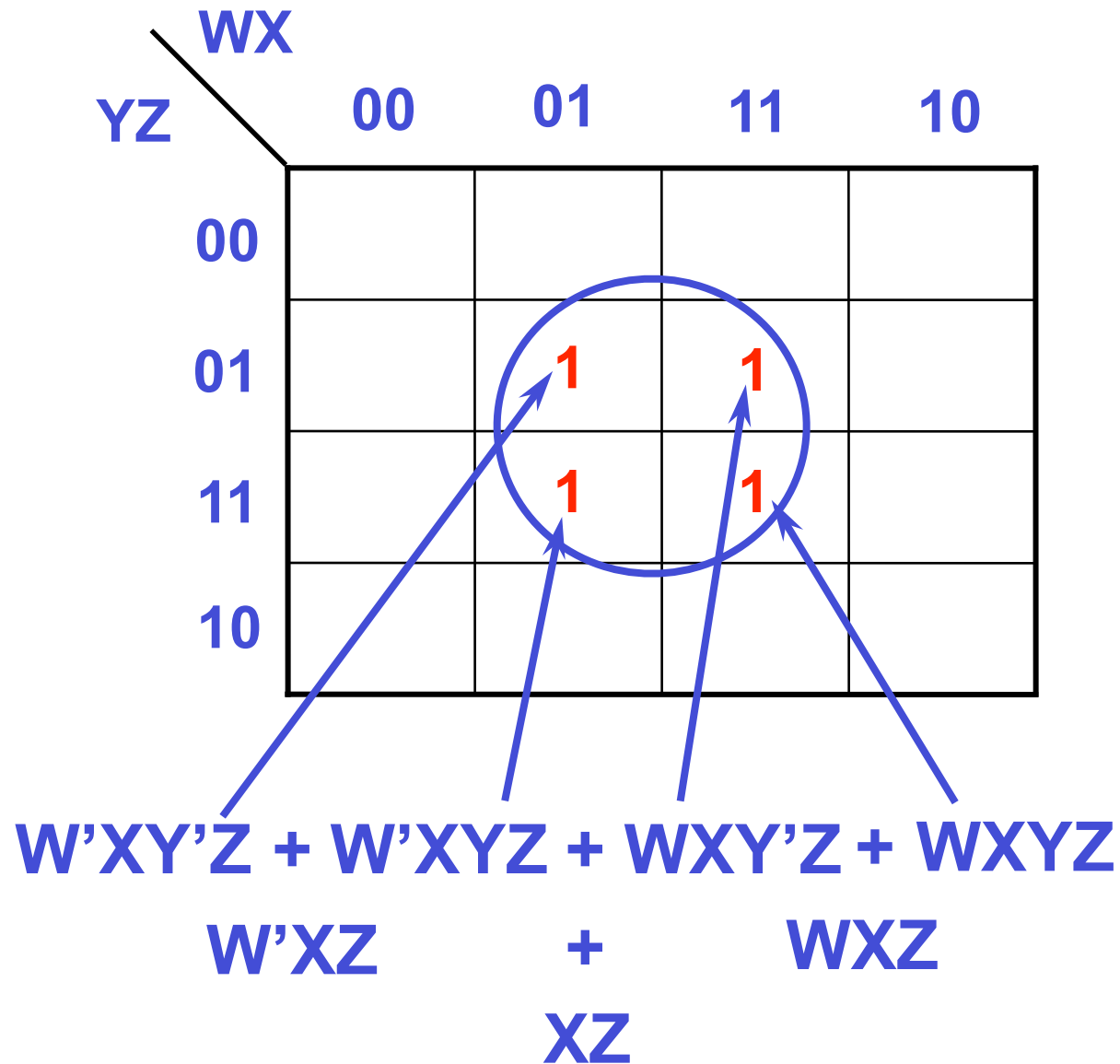
Minimization Using Karnaugh Map

- **Goal: Cover all 1-cells with the minimum number of Prime Implicants**
- **Procedure**
 - Plot 1's corresponding to minterms of function
 - Circle largest possible rectangular sets of 1's
 - Must be power of 2
 - Including "wrap-around" sets
 - Repeat until all minterms are covered
 - OR product terms derived from each circle
- **Minimizes number of gates and gate inputs**

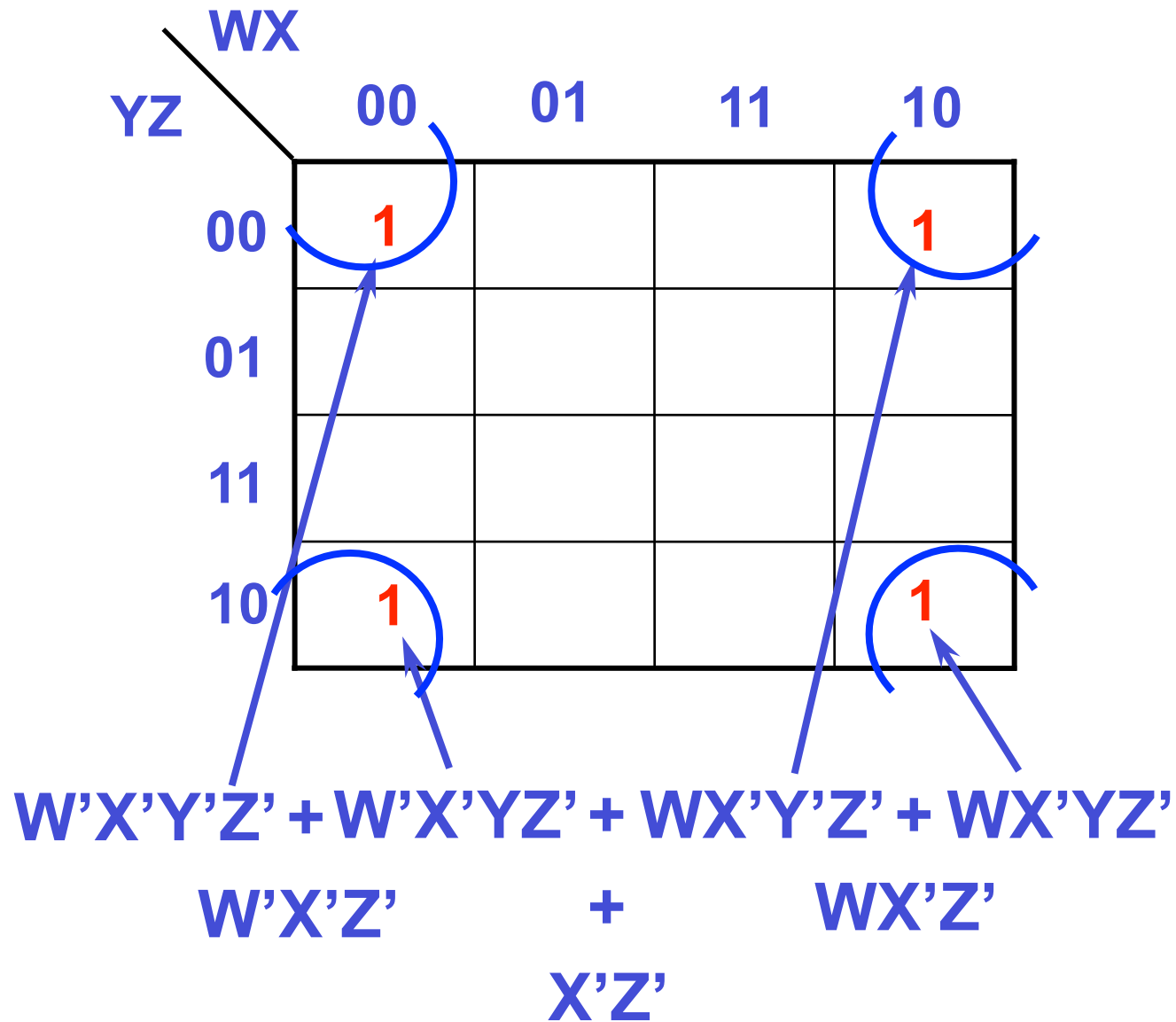
4 Variable Karnaugh Map

		WX			
		00	01	11	10
YZ	00				
	01				
	11				
	10				

Combining Theorem in Action



Combining Theorem in Action



3 Variable Karnaugh Map Example

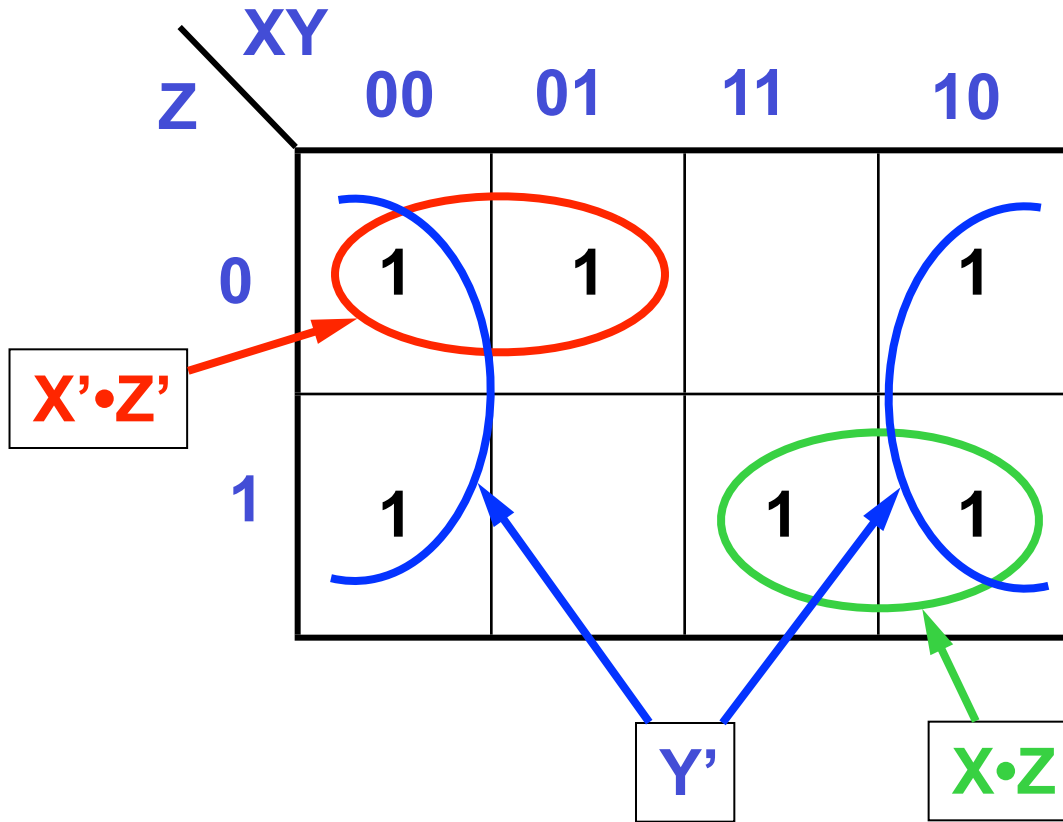
$$F = \sum_{X,Y,Z} (4,5,6,7)$$

Z \ XY	00	01	11	10
0			1	1
1			1	1

$$F = X$$

Another Example

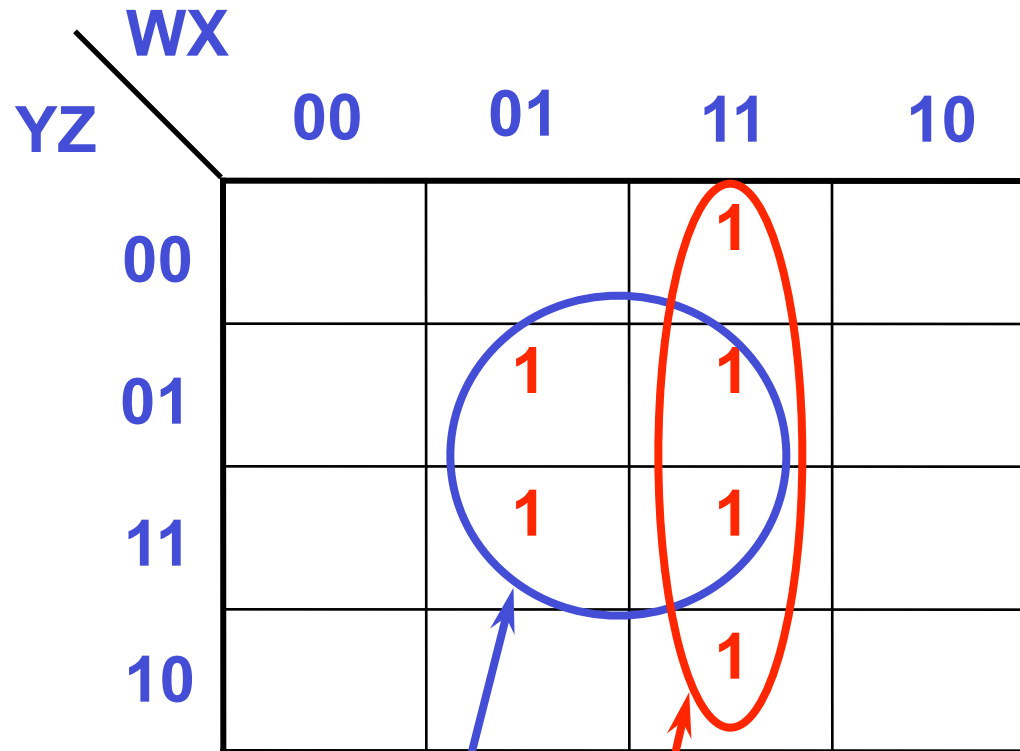
$$F = \sum_{x,y,z}(0,1,2,4,5,7)$$



$$F = X' \cdot Z' + Y' + X \cdot Z$$

4 Variable Karnaugh Map Example

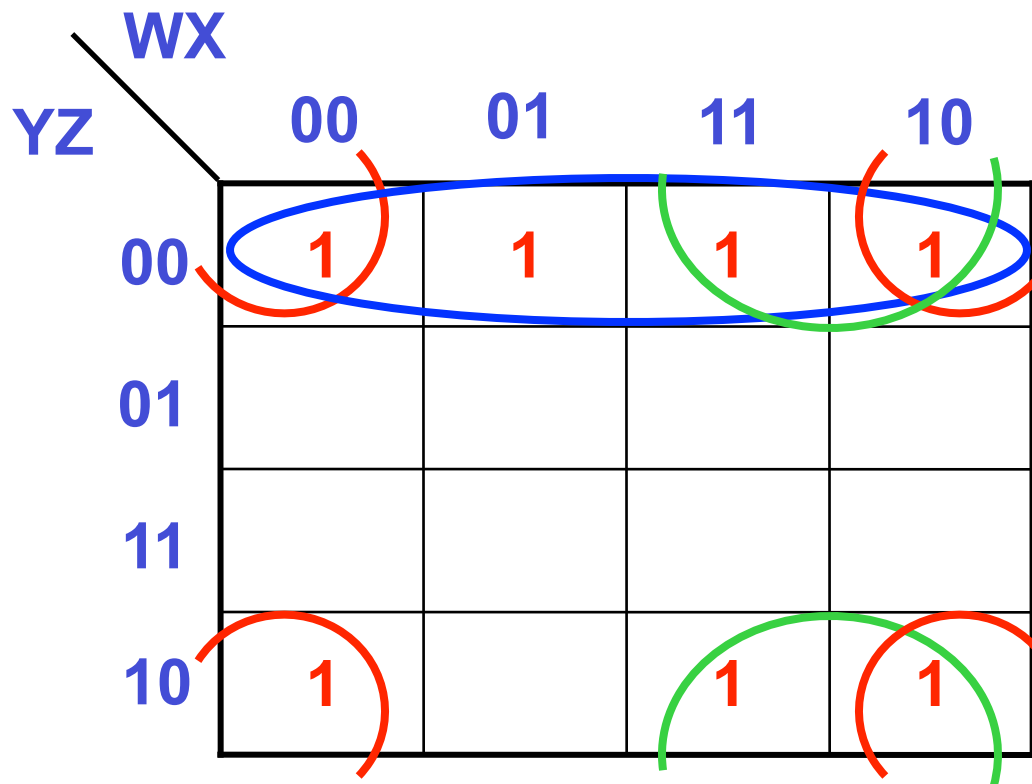
$$F = \sum_{W,X,Y,Z} (5,7,12,13,14,15)$$



$$F = X \cdot Z + W \cdot X$$

Another Example

- Detect all even digits from 0 to 15 except 6
 - $F = \Sigma_{W,X,Y,Z}(0,2,4,8,10,12,14)$



$$F = X'Z' + WZ' + Y'Z'$$

Yet Another

- Minimize $F = \sum_{A,B,C,D}(0, 1, 4, 5, 6, 7, 9, 13, 14, 15)$

		AB			
		00	01	11	10
CD	00	1	1		
	01	1	1	1	1
	11		1	1	
	10		1	1	

$$F = A' \cdot C' + B \cdot C + C' \cdot D$$

Minimizing Product-of-Sums

- **Procedure**

- Plot 0's corresponding to maxterms of function
- Circle largest possible rectangular sets of 0's
 - Must be power of 2
 - Including “wrap-around” sets
- Repeat until all maxterms are covered
- AND sum terms derived from each circle

Product-of-Sums Example

$$F = \prod_{w,x,y,z} (0, 2, 6, 7, 8, 10)$$

		WX			
		00	01	11	10
YZ	00	0			0
	01				
	11		0		
	10	0	0		0

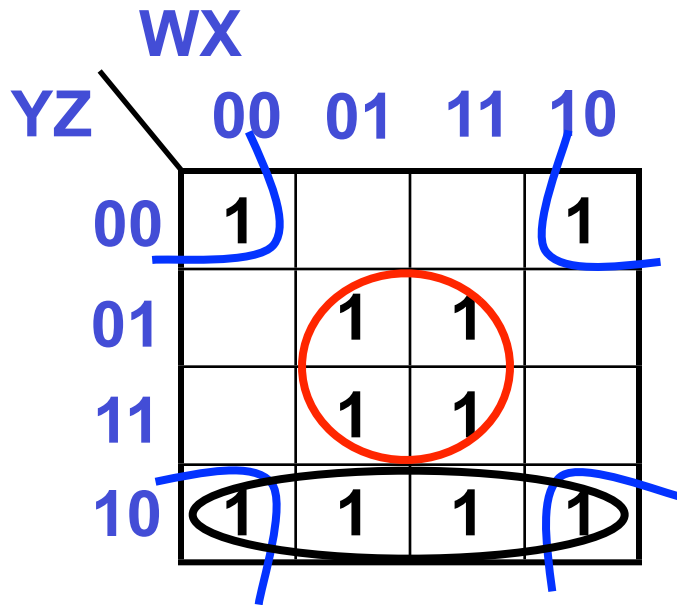
Corners:
 $X+Z$

Other:
 $W+X'+Y'$

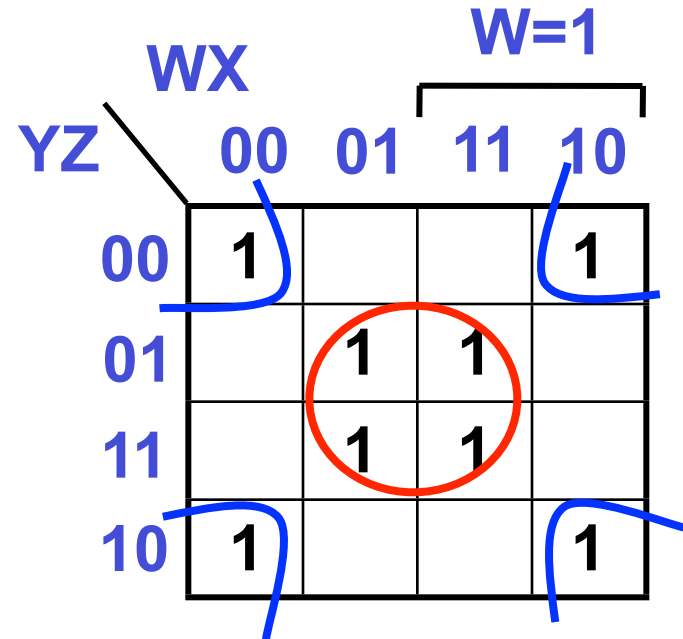
$$F = (X+Z) \cdot (W+X'+Y')$$

5 Variable Karnaugh Maps

$$F = \sum_{VWXYZ} (0, 2, 5, 6, 7, 8, 10, 13, 14, 15, 16, 18, 21, 23, 24, 26, 29, 31)$$



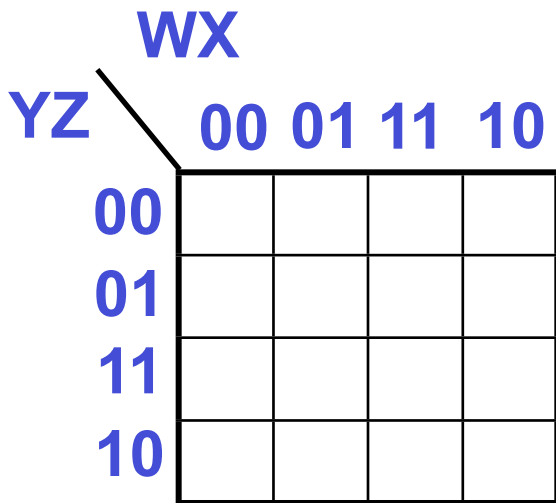
V=0



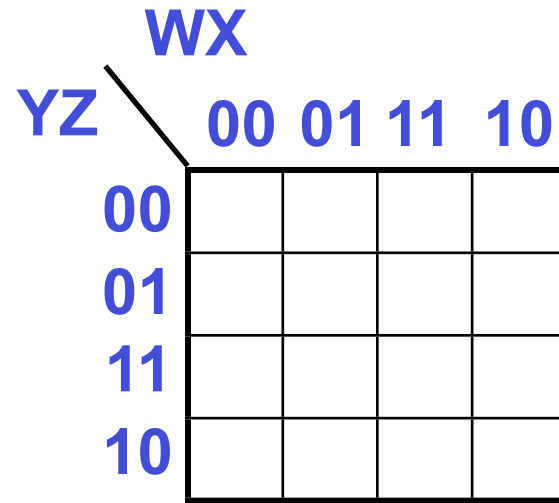
V=1

$$F = X' \cdot Z' + X \cdot Z + V' \cdot Y \cdot Z'$$

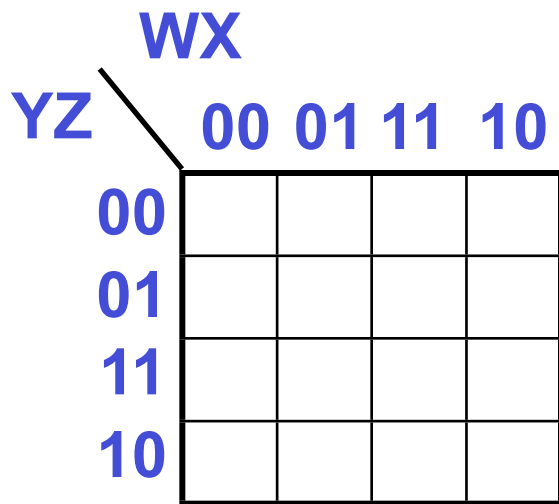
6 Variable Karnaugh Maps



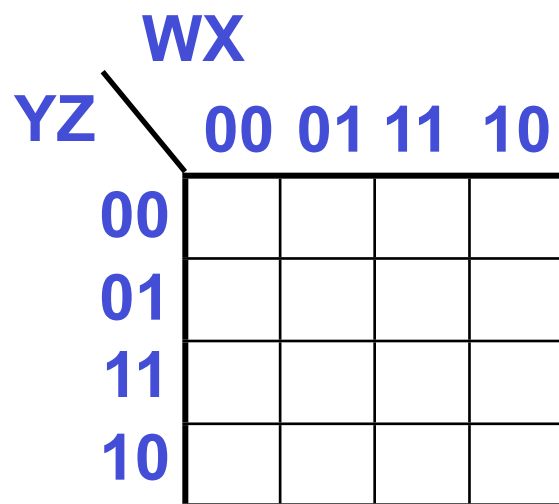
$U,V=0,0$



$U,V=0,1$



$U,V=1,0$



$U,V=1,1$

Don't-Care Combinations

- **Sometimes the output for a particular input combination is irrelevant**
 - **Such as an input combination that will never happen**
- **Can be used as 1- or 0-cells as needed**
- **Represent as a 'd' in the Karnaugh Map**
- **Only circle if doing so creates a larger Prime Implicant (and thus a more minimal expression)**

Don't-Care Example

- Detect all even decimal digits except 6
- Only inputs 0-9 appear
- 10-15 are “don't care” values

YZ \ WX	00	01	11	10
00	1	1	d	1
01			d	
11			d	d
10	1		d	d

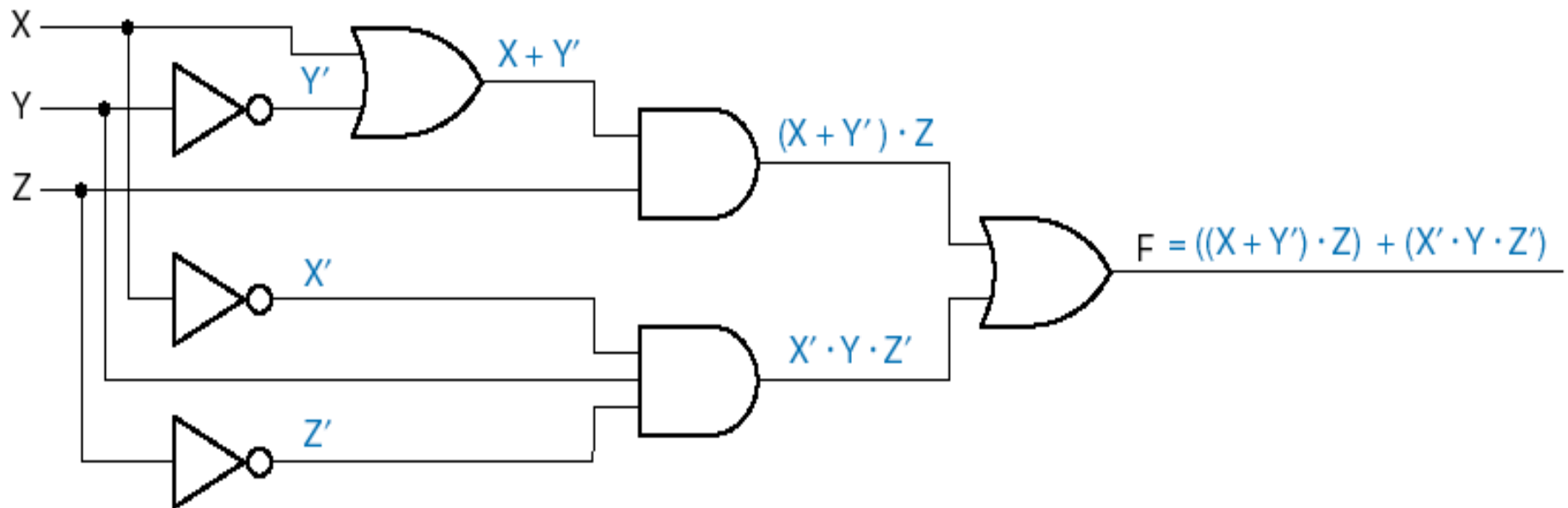
$$F = Y' \cdot Z' + X' \cdot Z'$$

Deriving a Circuit Output Function

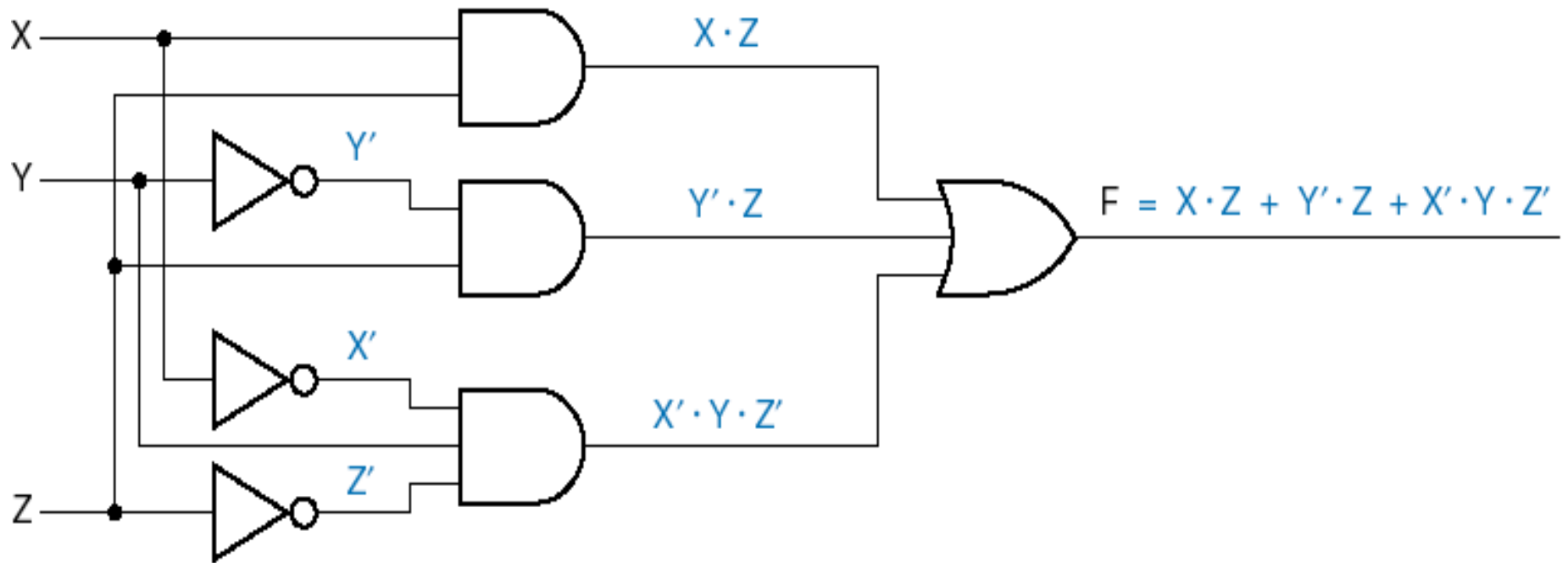
Start from the inputs

Derive intermediate outputs

Carry through to the circuit output

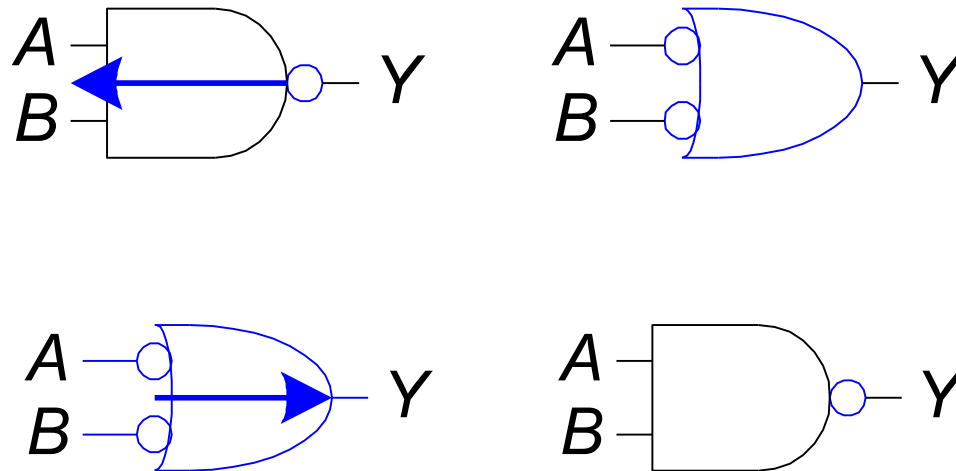


Deriving a Circuit Output Function



Bubble Pushing

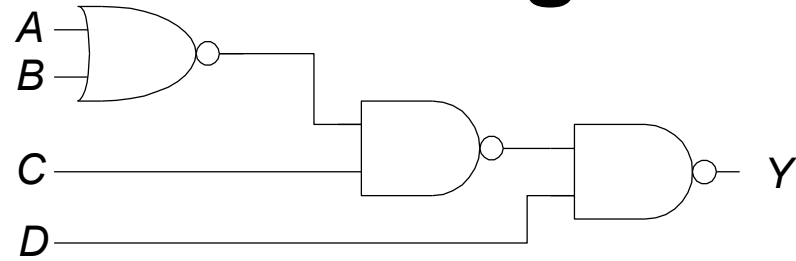
- Aid for deriving circuit function
- Makes use of De Morgan to push bubbles from output to inputs, or vice-versa



Bubble Pushing Rules

- **Work from output back to inputs**
- **Push bubble on final output back**
- **Draw gates in a form so that bubbles cancel**

Bubble Pushing Example



$$Y = A'B'C + D'$$

Before Next Class

- H&H 1.7, 2.8

Next Time

CMOS Logic
Logic Functions