

ECE 5330: Homework #6 Solutions

Problem 6.1

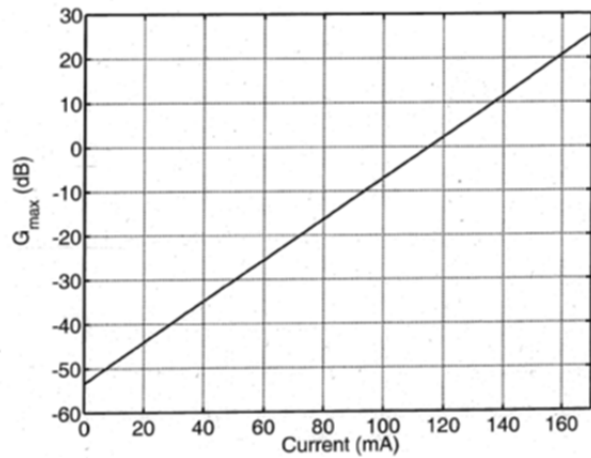
a) Use the slab waveguide approximation:

$$n_g n_{eff} = \frac{\iint n n_g^M |\phi|^2 dx dy}{\iint |\phi|^2 dx dy} \quad \Rightarrow \quad n_g = 3.2355$$

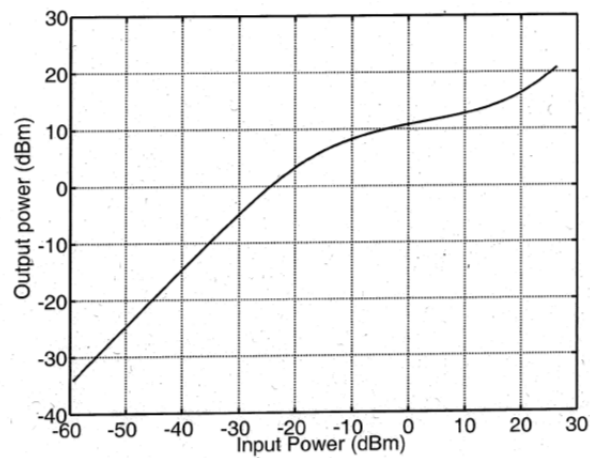
b) $A_{eff} = \frac{A_a}{\Gamma_a} = 1.5 (\mu m)^2$

c) $P_{sat} = \frac{\hbar \omega A_{eff}}{\tilde{a}_o \tau_r} = 6 \text{ mW}$

d) $G^* = e^{(\Gamma_a \tilde{g}^* - \tilde{\alpha})L}$ and $\Gamma_a \tilde{g}^* = \Gamma_a \tilde{a}_o (n(z) - n_{tr}) \big|_{P(z) \ll P_{sat}} = \Gamma_a \tilde{a}_o \left(\frac{\eta_i I \tau_r}{q V_a} - n_{tr} \right)$. See the plot below.



e) $\Gamma_a \tilde{g}^* / \tilde{\alpha} = 4.88$ and $G^* = e^{(\Gamma_a \tilde{g}^* - \tilde{\alpha})L} \approx 341$.

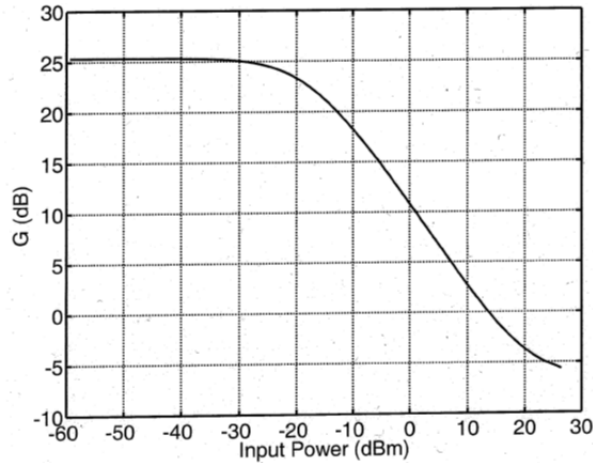


I have used

$$\frac{P(0)}{P_{sat}} = \left(\frac{\Gamma_a \tilde{g}^*}{\tilde{\alpha}} - 1 \right) \left(\frac{e^{\frac{(\Gamma_a \tilde{g}^* - \tilde{\alpha})L - \ln G}{\Gamma_a \tilde{g}^* / \tilde{\alpha}}} - 1}{G e^{\frac{(\Gamma_a \tilde{g}^* - \tilde{\alpha})L - \ln G}{\Gamma_a \tilde{g}^* / \tilde{\alpha}}} - 1} \right) = \left(\frac{\Gamma_a \tilde{g}^*}{\tilde{\alpha}} - 1 \right) \left(\frac{\left(\frac{G^*}{G} \right)^{\tilde{\alpha} / \Gamma_a \tilde{g}^*} - 1}{G \left(\frac{G^*}{G} \right)^{\tilde{\alpha} / \Gamma_a \tilde{g}^*} - 1} \right)$$

The easiest way to do this in matlab is to assume values for G and then calculate $P(0)$ using the above expression, and then plot the graph.

f)



g) Use,

$$P_{ASE} = \left(\frac{\Delta \omega_f}{2\pi} \right) \hbar \omega_f \frac{\Gamma_a \tilde{g}(\omega_f)}{\Gamma_a \tilde{g}(\omega_f) - \tilde{\alpha}} n_{sp}(\omega_f) \left(e^{(\Gamma_a \tilde{g}(\omega_f) - \tilde{\alpha})L} - 1 \right) \approx 51 \mu W$$

Problem 6.2

a) Integrate,

$$\frac{dP(z)}{dz} = \left[\frac{\Gamma_a \tilde{g}^*}{1 + \frac{P(z)}{P_{sat}}} - \tilde{\alpha} \right] P(z)$$

from 0 to z to get,

$$\frac{P(0)}{P_{sat}} = \left(\frac{\Gamma_a \tilde{g}^*}{\tilde{\alpha}} - 1 \right) \left(\frac{e^{\frac{(\Gamma_a \tilde{g}^* - \tilde{\alpha})L - \ln G(z)}{\Gamma_a \tilde{g}^* / \tilde{\alpha}}} - 1}{G(z) e^{\frac{(\Gamma_a \tilde{g}^* - \tilde{\alpha})L - \ln G(z)}{\Gamma_a \tilde{g}^* / \tilde{\alpha}}} - 1} \right) = \left(\frac{\Gamma_a \tilde{g}^*}{\tilde{\alpha}} - 1 \right) \left(\frac{\left(\frac{G^*}{G(z)} \right)^{\tilde{\alpha} / \Gamma_a \tilde{g}^*} - 1}{G(z) \left(\frac{G^*}{G(z)} \right)^{\tilde{\alpha} / \Gamma_a \tilde{g}^*} - 1} \right)$$

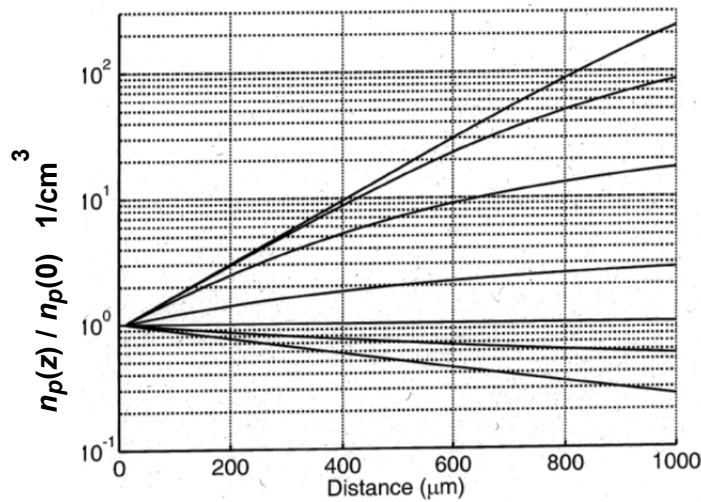
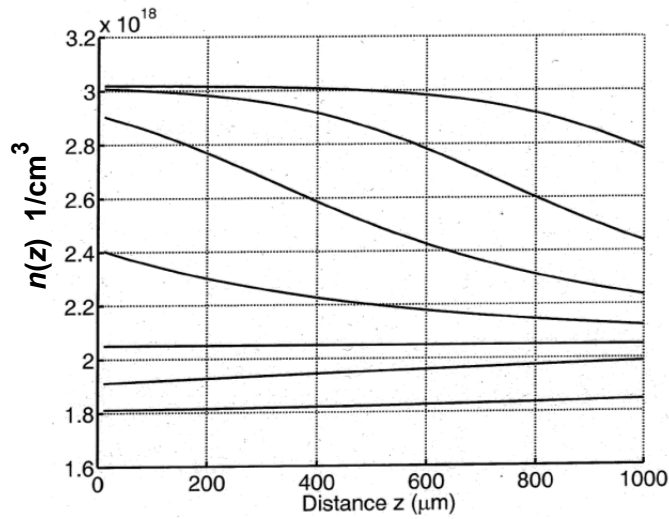
where,

$$G(z) = \frac{P(z)}{P_{sat}}$$

The above equation can be used to calculate $P(z)$ given $P(0)$ and then the carrier density can be obtained using,

$$n(z) - n_{tr} = \frac{\frac{\eta_i I \tau_r}{q V_a} - n_{tr}}{1 + \frac{P(z)}{P_{sat}}}$$

The results are plotted below.



b) When $P(0) \rightarrow \infty$, $n(z) \rightarrow n_{tr}$ and the amplifier is completely saturated.

c) When the input power is zero,

$$n(z) = \frac{\frac{\eta_i I \tau_r}{q V_a} - n_{tr}}{1 + \frac{P(z)}{P_{sat}}} + n_{tr} = \frac{\eta_i I \tau_r}{q V_a} = 3 \times 10^{18} \text{ 1/cm}^3$$

d) Perhaps a poorly worded question. The question is really asking, “What input power level will make the amplifier gain G unity and, therefore, the modal gain at every location in the amplifier equal to the modal loss at that location?”. If the modal gain is equal to the modal loss at every location, then this means that the carrier density $n(z)$ and the optical power $P(z)$ are both independent of position. This also means that, $P(L) = P(z) = P(0)$. If $G = 1$, we get,

$$\frac{P(0)}{P_{sat}} = \left(\frac{\Gamma_a \tilde{g}^*}{\tilde{\alpha}} - 1 \right)$$

e) This is the same situation as in part (d). When the gain of the amplifier is unity,

$$\Gamma_a \tilde{a}_o (n(z) - n_{tr}) = \tilde{\alpha}$$

$$\Delta n(z) = 2.05 \times 10^{18} \text{ 1/cm}^3$$