

ECE 4880

Fall 2016

Chapter 9

Phase Noises in the Signal Chain

Edwin C. Kan

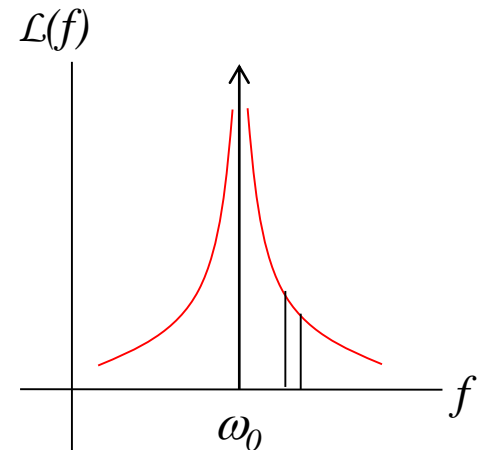
School of Electrical and Computer Engineering
Cornell University

Goals

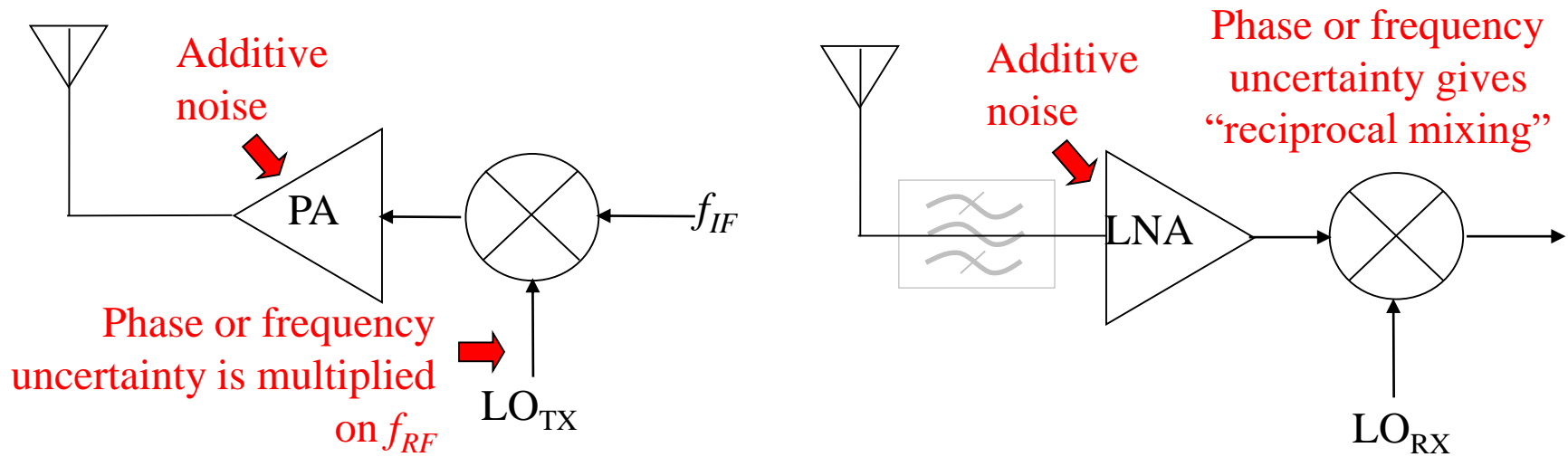
- Effects of amplitude and phase noises in RF transceivers
- Intuitive modeling of phase noises by LC resonators
- Empirical models for phase noises

Phase Noise Definition

- Noise free signal: $v(t) = A\cos(\omega_0 t)$
- Amplitude noise (AM): stochastic process by : $v(t) = A(t)\cos(\omega_0 t)$
- Phase noise (PM): stochastic process by: $A\cos(\omega_0 t + \varphi(t))$
- The phase noise is often represented as one-sided spectral density of a signal's phase deviation or phase instability $S_\varphi(f)$
- Phase noise is often defined by $\mathcal{L}(f)$ (as script L of f), which represents the noise power relative to the carrier contained in a 1Hz bandwidth centered at the carrier frequency.
- $\mathcal{L}(f)$ has the unit of dBc/Hz.
- For a $1/f$ phase noise, $\mathcal{L}(f)$ can be -80dBc/Hz at 10kHz offset or -90dBc/Hz at 100kHz.

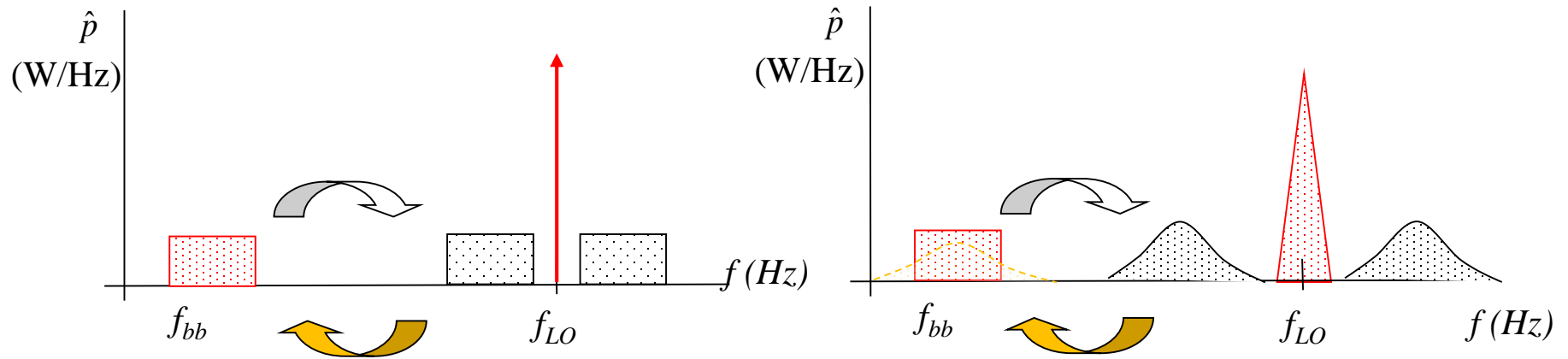


Phase Noise in Oscillators



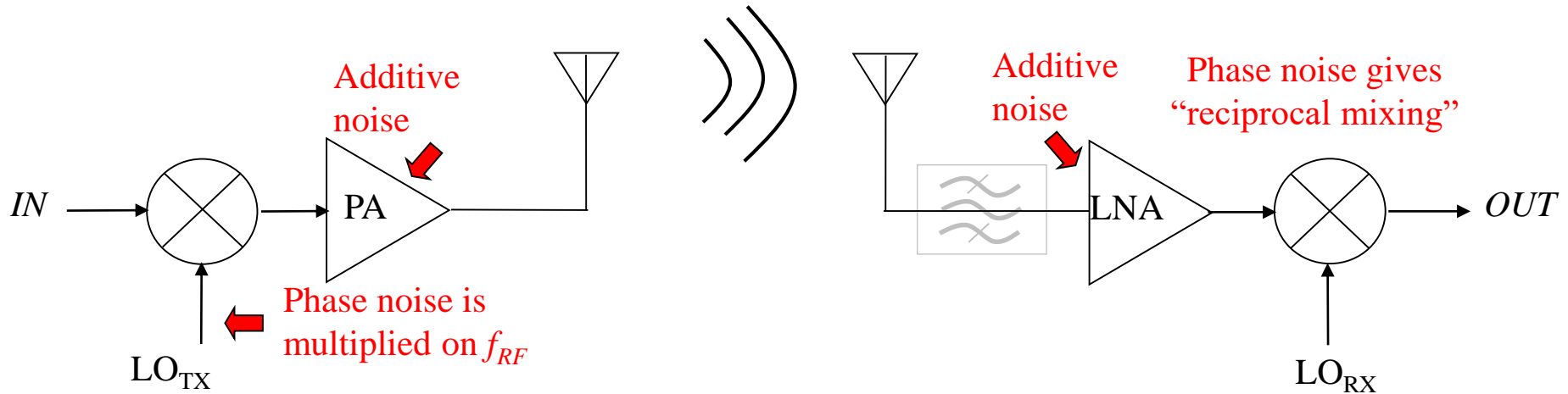
- Phase noises of LO will be multiplied to the signals in TX and RX (convolution in the frequency domain).
- LO_{TX} and LO_{RX} uncertainty will demand wider channel separation and larger separation of “symbols”.
- Silver lining: the phase shift offset between LO_{TX} and LO_{RX} carries the time-of-flight information.

Phase Noise in Oscillators



- Perfect LO as a delta function in the spectrum will just perform frequency translation with NO additional components.
- Phase noise will convolute into the mixer output as amplitude (envelope) and phase (zero crossing) noises, generally termed as “reciprocal mixing” of AM and PM.
- Frequency (FM) and phase (PM) modulation can be different for long integration cycles.
- Phase noise is especially serious for timing-based modulation.
- Phase noise has been important for any clocks.

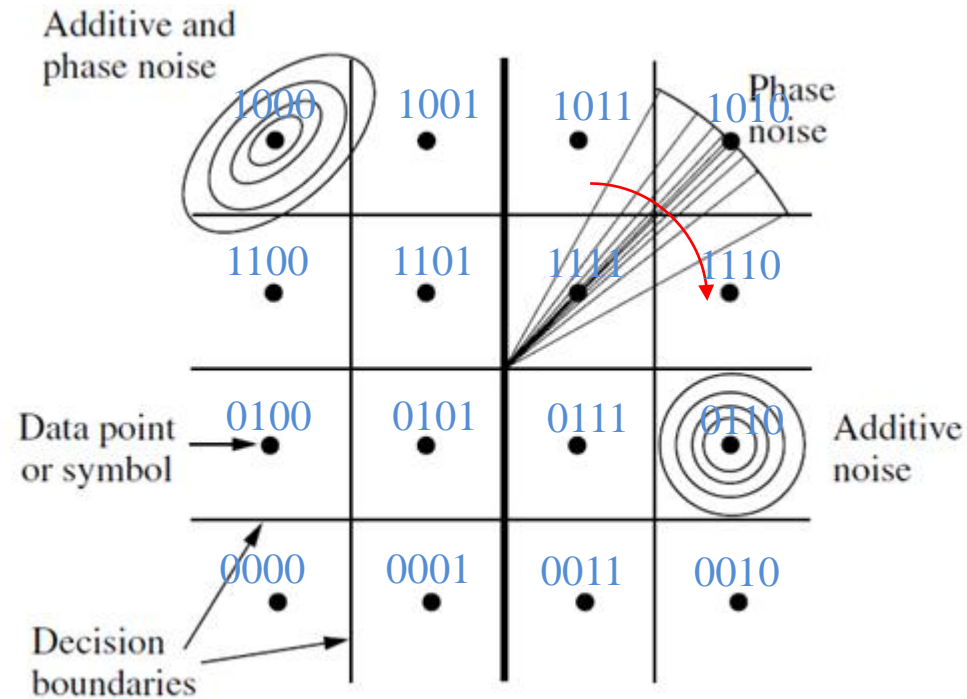
Wireless Link: Noises and Distortion



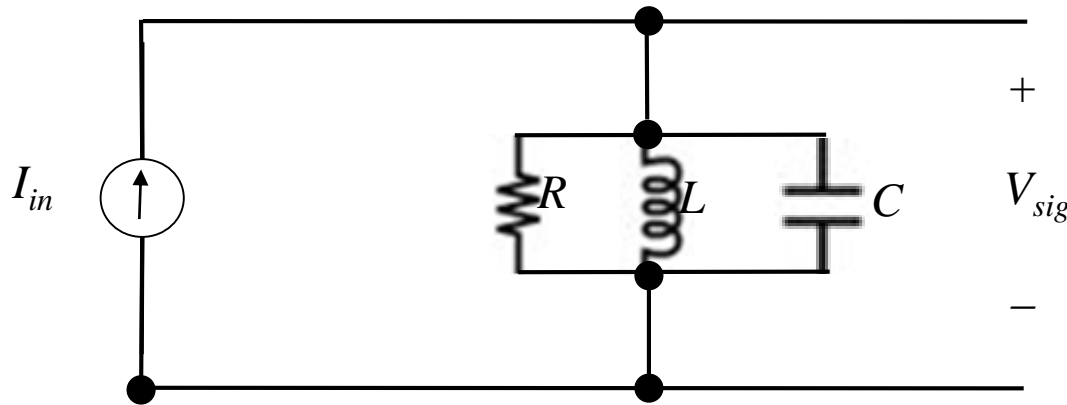
- The main purpose of wireless link is to retrieve information from IN to OUT , with the noise and nonlinearity of
 - Two mixers
 - Two LO generators
 - PA and LNA
 - Antenna
 - Free-space path

Effects of LO Phase Noise

- Receiver desensitization by the phase noises of LO_{TX} and LO_{RX}
- Jitter is the uncertainty in the synchronization (zero-crossing points) source:
 - AM: from uncertainty in loop gain (should be 1)
 - PM: from uncertainty in phase delay (should be $2n\pi$)
- Larger bit error rate in Q-ary modulation by phase error: only bordering domain has 1-bit error (easier for ECC).



Phase Noises in LC Oscillators

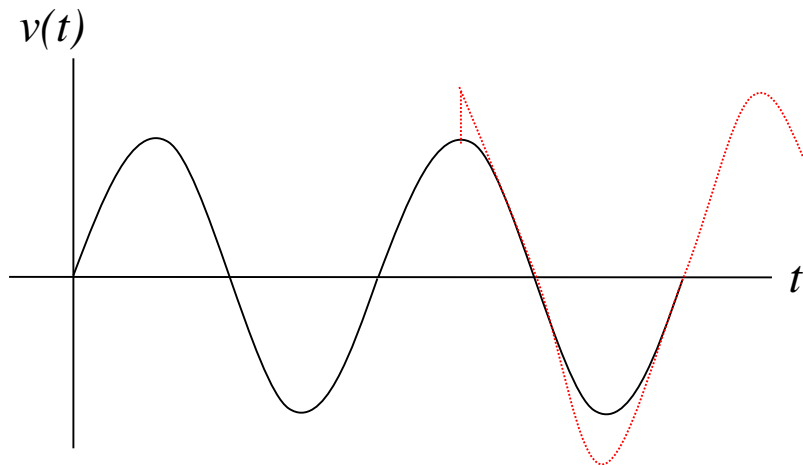


Stable solution: $V_A \cos(\omega_0 t)$

$$\text{Average } E_{\text{stored}}: \frac{CV_A^2}{2} \omega_0$$

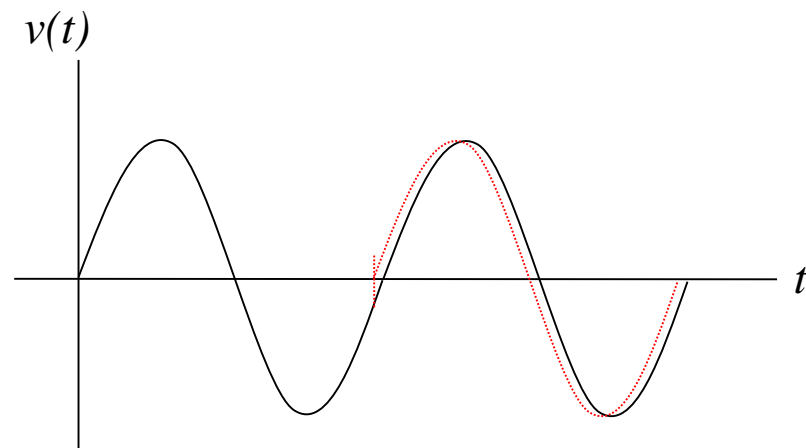
- LC resonators can serve as an oscillator when R is compensated by an active I_{in} : electrostatic and magnetostatic energies balanced each other
- Colpitts or Clapp or cross-coupled oscillators
- Piezoelectric resonators
- Mechanical oscillators (pendulum with compensation): AM and PM can be easiest visualized.
- Loop gain: 1; phase loop delay: $2n\pi$.

Energy Injection for AM and PM



Injection at peak :

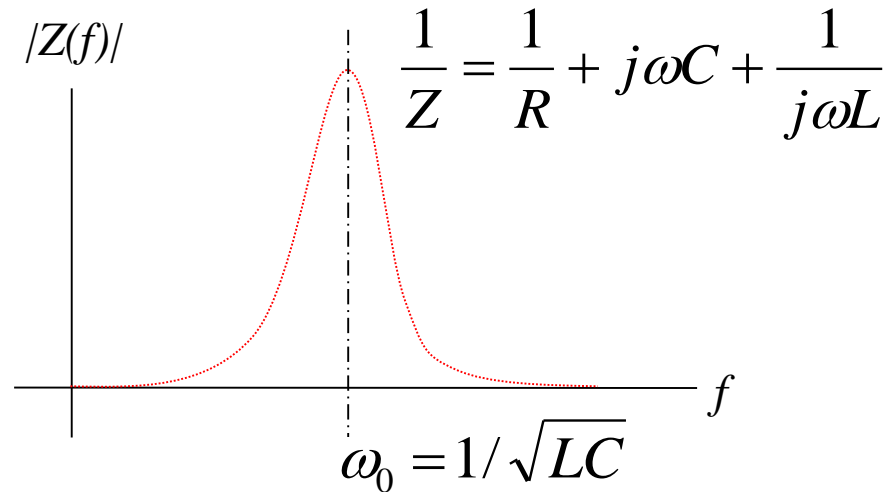
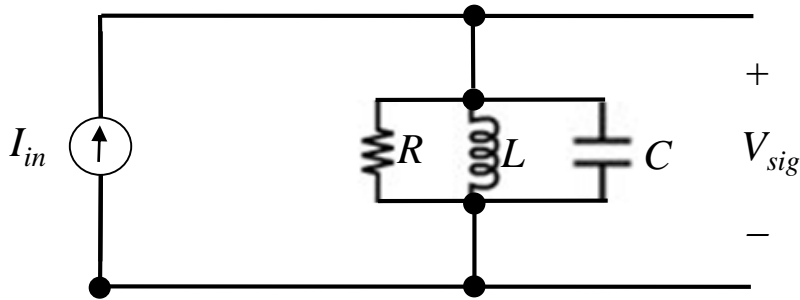
- Additional energy of $V_A \Delta I_{in}$
- Follow the solution of a higher amplitude (remember all amplitudes are allowed here)
- Zero crossing point will not change: only AM; no PM



Injection at zero crossing:

- No added energy
- Stay at the same magnitude V_A and frequency ω_0
- The local addition of charge will help the voltage reach its maximum faster: only PM; no AM!

Noises in the RLC Resonator



$$\bar{V}_n^2 = 4kTR \cdot BW = 4kTR \cdot \int_0^\infty \left| \frac{Z(f)}{R} \right|^2 df = 4kTR \cdot \frac{1}{4RC} = \frac{kT}{C}$$

$$Q \equiv \frac{P_{stored}}{P_{diss}} = \frac{\omega_0 E_{sig}}{P_{diss}} = \frac{R}{\omega_0 L} = \omega_0 RC$$

- R is the only noise source
- From 2nd law of thermodynamics, the noise energy is unretrievable, i.e., energy conservation system does not have noise

Noises Normalized to the Signal Carrier

- The energy stored in L and C is: $E_{sig} = C\bar{V}_{sig}^2$
- The noise-to-signal power ratio in an oscillator is (in dBc):

$$\frac{P_{noise}}{P_{signal}} = \frac{\bar{V}_n^2}{\bar{V}_{sig}^2} = \frac{kT}{E_{sig}}$$

- If the noise in R is equally partitioned in AM and PM:

$$P_{phase_noise} = \frac{kT}{2E_{sig}}$$

- The phase noise will not be white, but focus around ω_0
- **No** clean analytical form is yet available.

A Common Derivation of $1/f^2$ for $\mathcal{L}(f)$

- Impedance: $\frac{1}{Z} = \frac{1}{R} + j\omega C + \frac{1}{j\omega L}$

- Close to $\omega_0 + \Delta\omega$:

$$|Z(\omega_0 + \Delta\omega)| \cong R \cdot \frac{\omega_0}{2Q\Delta\omega} \quad \text{Im}(Z(\omega_0 + \Delta\omega)) \cong -j \cdot \frac{\omega_0 L}{2 \left(\frac{\Delta\omega}{\omega_0} \right)}$$

- If the total phase noise power is from the **thermal noise**:

$$P_{\text{phase_noise}} = \frac{kT}{2E_{\text{sig}}} = 2 \int_{\omega_0}^{\infty} df \mathcal{L}(f)$$

$$\text{We must have: } \mathcal{L}(\Delta\omega) = \frac{2kT}{P_{\text{sig}}} \cdot \left(\frac{\omega_0}{2Q\Delta\omega} \right)^2$$

Empirical Formula *for* $\mathcal{L}(f)$

- Leeson's approximation:

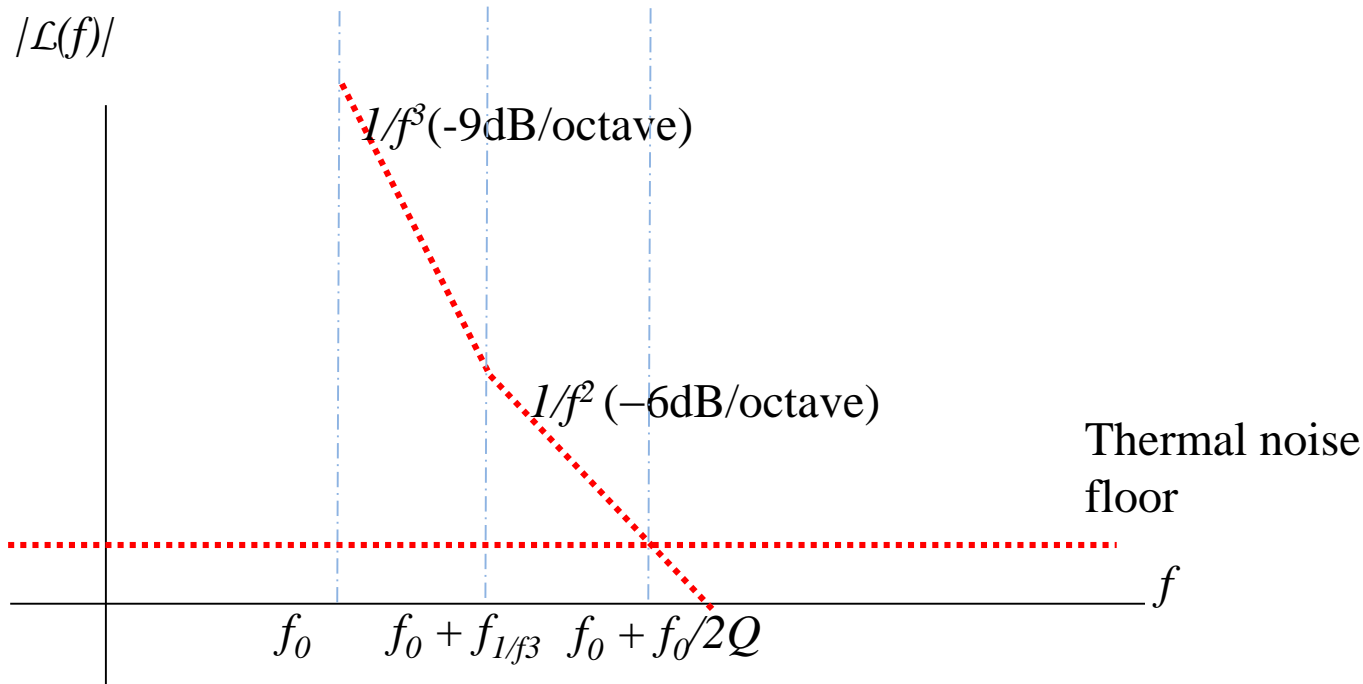
$$\mathcal{L}(\Delta\omega) = \frac{2FkT}{P_{sig}} \cdot \left[1 + \left(\frac{\omega_0}{2Q\Delta\omega} \right)^2 \right] \cdot \left(1 + \frac{\Delta\omega_{1/f^3}}{|\Delta\omega|} \right)$$

- Parameters: Q and $\Delta\omega_{1/f^3}$
- Thermal noise and Flicker noise contributions

Leeson's Approximation for $\mathcal{L}(f)$

- $f_0/2Q > f_{1/f^3}$ (higher thermal noise and lower Flicker noise)

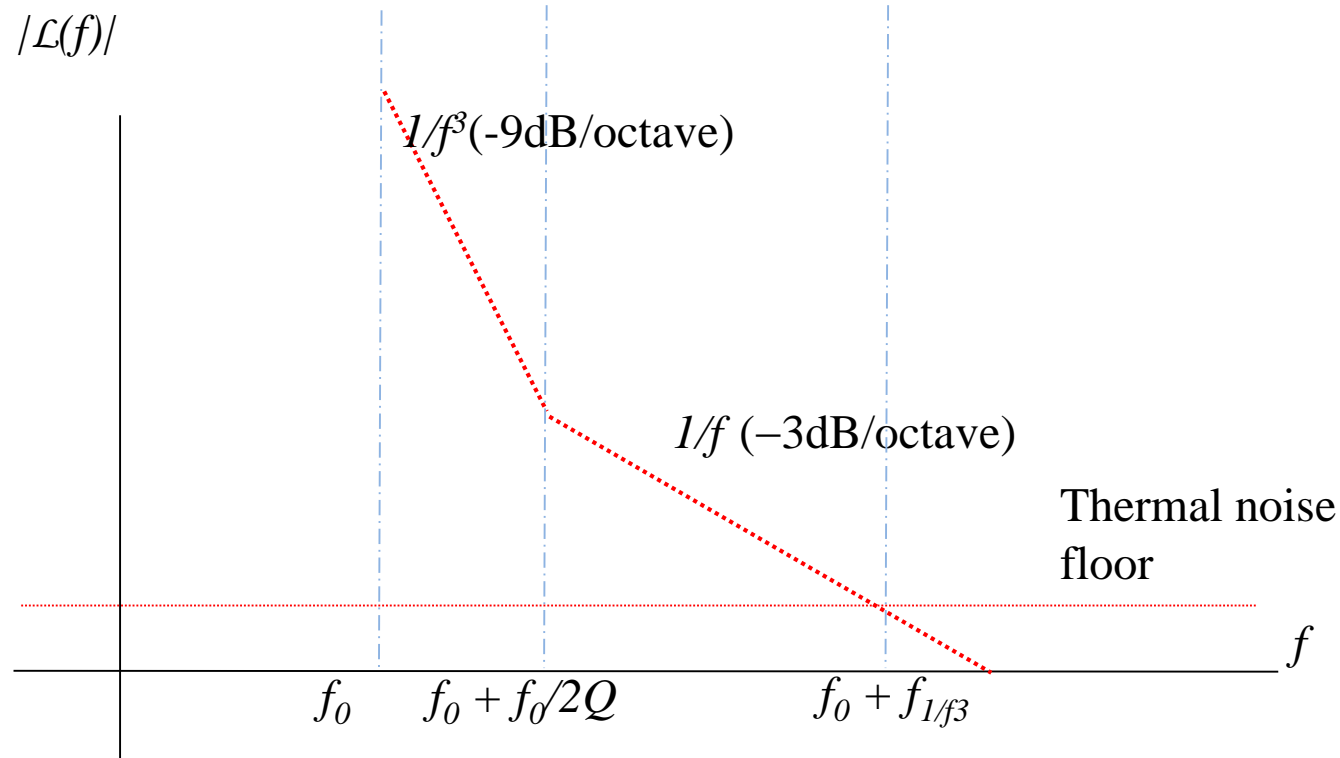
$$\mathcal{L}(\Delta\omega) = \frac{2FkT}{P_{sig}} \cdot \left[1 + \left(\frac{\omega_0}{2Q\Delta\omega} \right)^2 \right] \cdot \left(1 + \frac{\Delta\omega_{1/f^3}}{|\Delta\omega|} \right)$$



Leeson's Approximation for $\mathcal{L}(f)$

- $f_0/2Q < f_{1/f^3}$ (lower thermal noise, i.e., large Q and higher Flicker noise)

$$\mathcal{L}(\Delta\omega) = \frac{2FkT}{P_{sig}} \cdot \left[1 + \left(\frac{\omega_0}{2Q\Delta\omega} \right)^2 \right] \cdot \left(1 + \frac{\Delta\omega_{1/f^3}}{|\Delta\omega|} \right)$$



What Have You Learned?

- The origin and effects of phase noises
- AM and PM distribution and effects
- Oscillator LC models
- Empirical PM models