

## Lecture 12

### Single Stage FET Amplifiers: Common Gate Amplifier Common Drain Amplifier

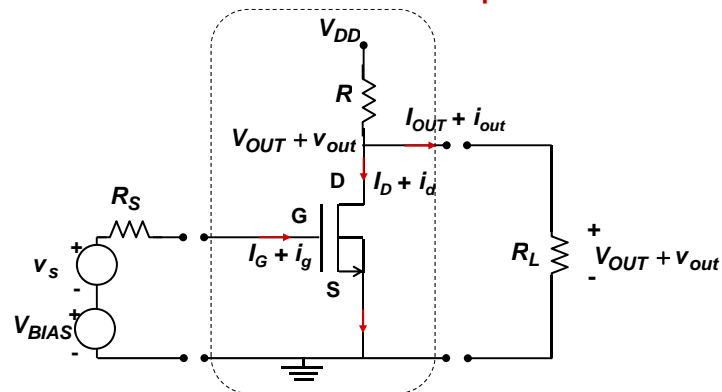
#### The Building Blocks of Analog Circuits - II

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In this lecture you will learn:

- Common Gate (CG) and Common Drain (CD) Amplifiers
- Small signal models of amplifiers

### The Common Source Amplifier



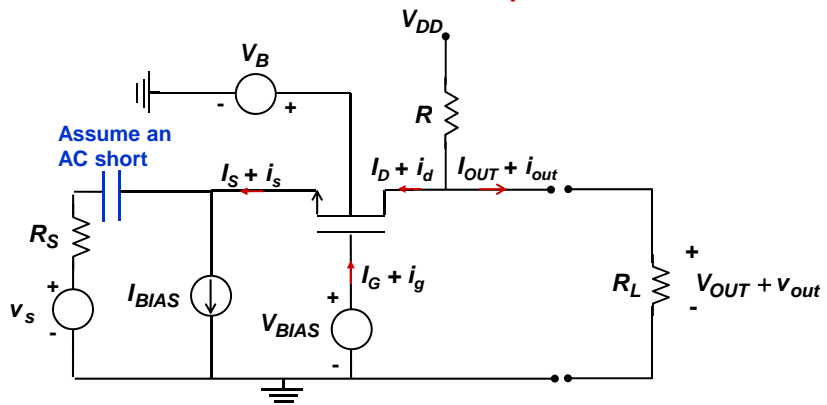
An attribute of the common source amplifier:

The input resistance is very large:

$$R_{in} = \infty \quad (\text{at least at low frequencies})$$

Not suitable for current amplifiers or transimpedance amplifiers

### The Common Gate Amplifier

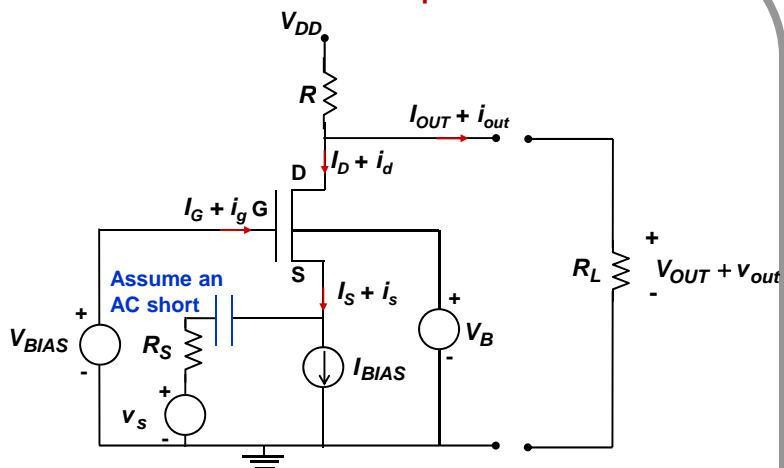


The gate terminal is “common” between the input and the output

The common gate amplifiers are useful when small input resistances and large output resistances are desired in current amplifiers

Note: The bulk is not tied to the source

### The Common Gate Amplifier

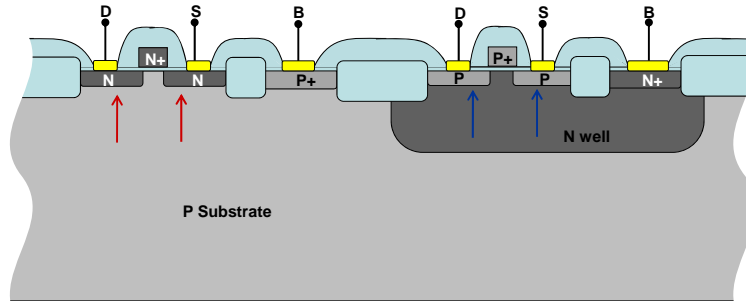


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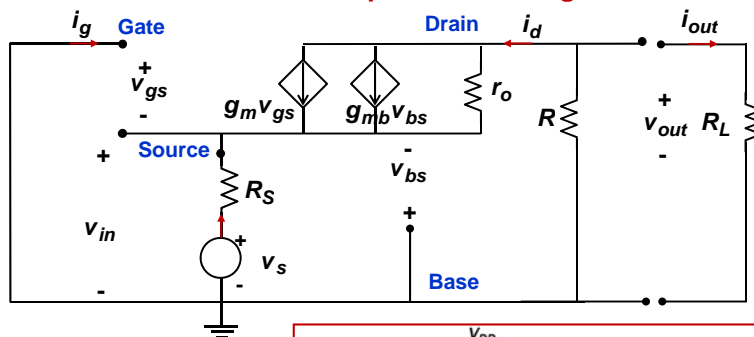
### Choice of Substrate Potentials



In order to keep the source and drain PN-junctions with the substrate (or bulk) reverse biased at all times,

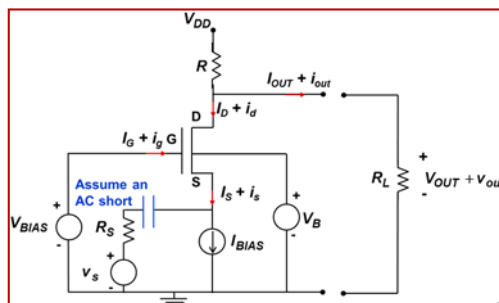
- i) The P-substrate (for NFETs) is generally tied to the most negative voltage in the circuit
- ii) The N-substrate or N-well (for PFETs) is tied to the most positive voltage in the circuit

### The Common Gate Amplifier: Small Signal Model

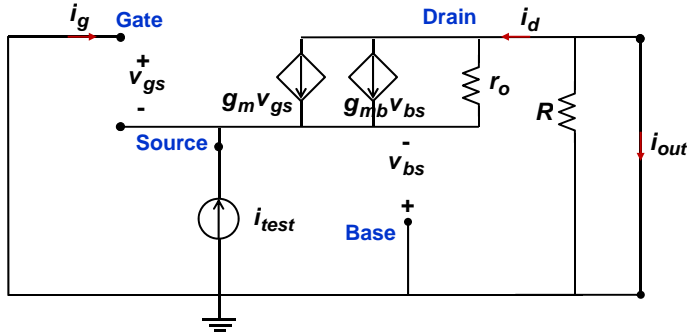


Note:

$$V_{bs} = V_{gs}$$



### The Common Gate Amplifier: Short Circuit Current Gain

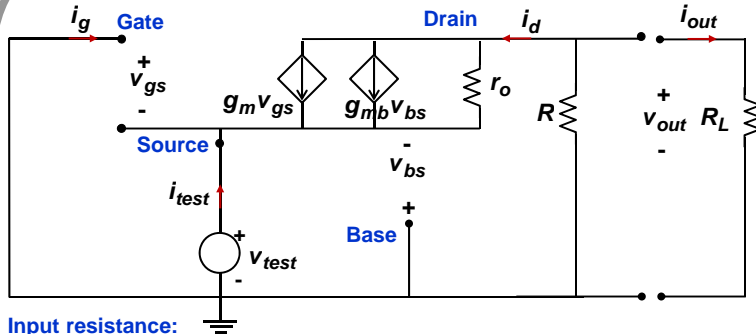


Short circuit current gain:

$$i_{test} = -i_d = i_{out}$$

$$A_i = \frac{i_{out}}{i_{test}} = 1 \longrightarrow \text{Short circuit current gain is unity!}$$

### The Common Gate Amplifier: Input Resistance



Input resistance:

$$i_{test} = -i_d = -\left(\frac{(g_m + g_{mb})r_o + 1}{r_o + (R \parallel R_L)}\right) v_{gs}$$

$$R_{in} = \frac{v_{test}}{i_{test}} = -\frac{v_{gs}}{i_{test}} = \frac{r_o + (R \parallel R_L)}{(g_m + g_{mb})r_o + 1}$$

$R_{in}$  can be small if:

$$(g_m + g_{mb})r_o \gg 1$$

$$r_o \gg (R \parallel R_L)$$

$$\Rightarrow R_{in} = \frac{r_o + (R \parallel R_L)}{(g_m + g_{mb})r_o + 1} \approx \frac{1}{(g_m + g_{mb})} \longrightarrow \text{Small}$$

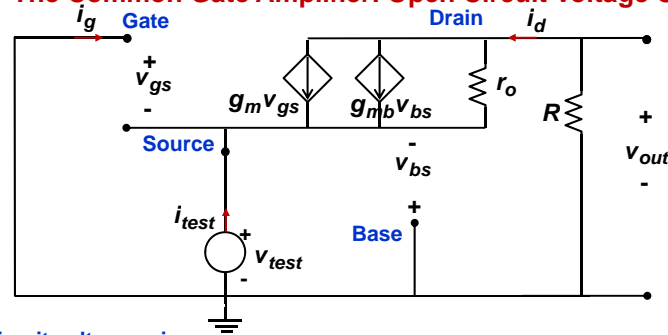
$$i_d = (g_m + g_{mb})v_{gs} + \frac{v_{out} + v_{gs}}{r_o}$$

$$= (g_m + g_{mb})v_{gs} + \frac{v_{gs}}{r_o} - i_d \frac{(R \parallel R_L)}{r_o}$$

$$i_d = \frac{(g_m + g_{mb})v_{gs} + \frac{v_{gs}}{r_o}}{1 + \frac{(R \parallel R_L)}{r_o}}$$

$$\approx \frac{(g_m + g_{mb})r_o + 1}{r_o + (R \parallel R_L)} v_{gs}$$

### The Common Gate Amplifier: Open Circuit Voltage Gain



Open circuit voltage gain:

$$v_{out} = -i_d R = -R \left( \frac{(g_m + g_{mb}) r_o + 1}{r_o + R} \right) v_{gs} = - \left( g_m + g_{mb} + \frac{1}{r_o} \right) (R \parallel r_o) v_{gs}$$

$$\Rightarrow A_v = \frac{v_{out}}{v_{test}} = - \frac{v_{out}}{v_{gs}} = \left( g_m + g_{mb} + \frac{1}{r_o} \right) (R \parallel r_o)$$

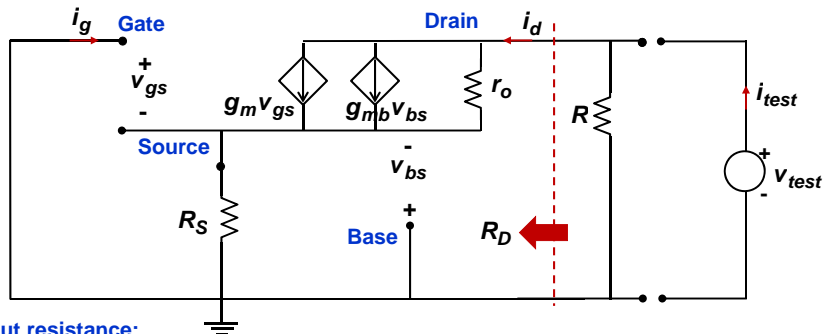
$$\left. \begin{aligned} i_d &= (g_m + g_{mb}) v_{gs} + \frac{v_{out} + v_{gs}}{r_o} \\ &= (g_m + g_{mb}) v_{gs} + \frac{v_{gs}}{r_o} - i_d \frac{R}{r_o} \\ i_d &= \frac{(g_m + g_{mb}) r_o + 1}{r_o + R} v_{gs} \end{aligned} \right\}$$

If:

$$(g_m + g_{mb}) r_o \gg 1$$

$$\Rightarrow A_v = \left( g_m + g_{mb} + \frac{1}{r_o} \right) (R \parallel r_o) \sim (g_m + g_{mb}) (R \parallel r_o) \rightarrow \text{Large if } R \text{ is large}$$

### The Common Gate Amplifier: Output Resistance



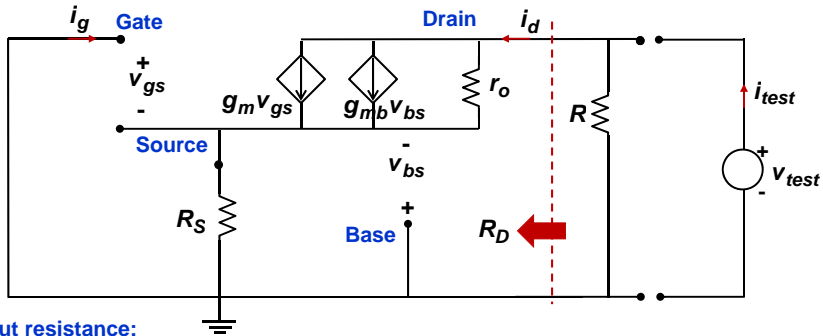
Output resistance:

$$i_d = (g_m + g_{mb}) v_{gs} + \frac{v_{test} + v_{gs}}{r_o} = - \frac{v_{gs}}{R_s}$$

$$\Rightarrow v_{gs} = -v_{test} \frac{\left( r_o \parallel \frac{1}{(g_m + g_{mb})} \parallel R_s \right)}{r_o}$$

$$R_D = \frac{v_{test}}{i_d} = \frac{r_o R_s}{\left( r_o \parallel \frac{1}{(g_m + g_{mb})} \parallel R_s \right)} = r_o + R_s [1 + r_o (g_m + g_{mb})]$$

### The Common Gate Amplifier: Output Resistance



Output resistance:

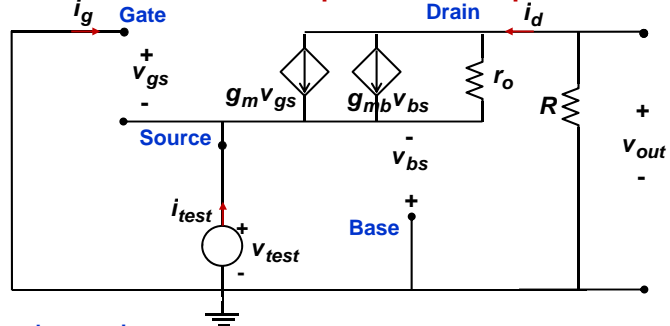
$$R_D = r_o + R_S + r_o R_S (g_m + g_{mb})$$

$$R_{out} = (R \parallel R_D) \\ = (R \parallel [r_o + R_S + r_o R_S (g_m + g_{mb})])$$

→  $R_{out}$  can be large if  $R$  is large

$R_{out}$  depends on  $R_S$   
(Non-unilateral)

### The Common Gate Amplifier: Transimpedance Gain



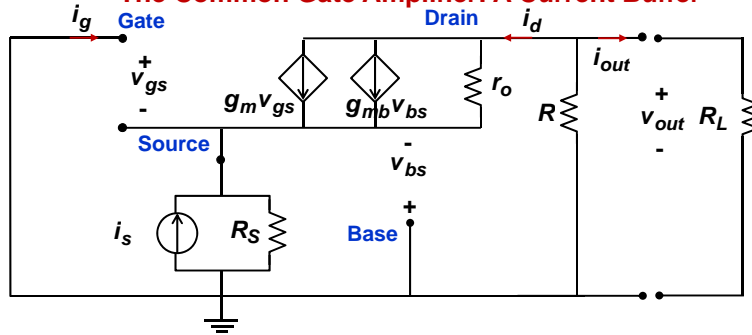
Transimpedance gain:

$$i_{test} = -i_d$$

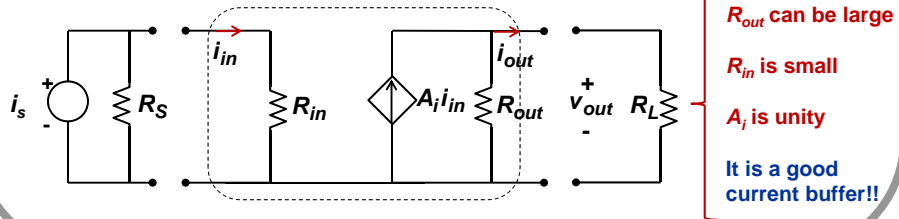
$$R_m = \frac{v_{out}}{i_{test}} = -\frac{i_d R}{i_{test}} = R$$

→  $R_m$  can be large if  $R$  is large

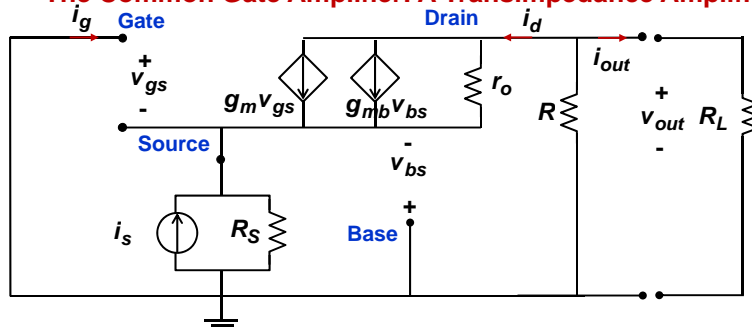
### The Common Gate Amplifier: A Current Buffer



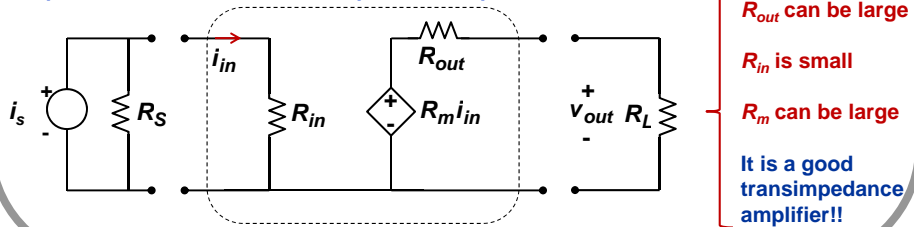
Compare with the standard current amplifier model:



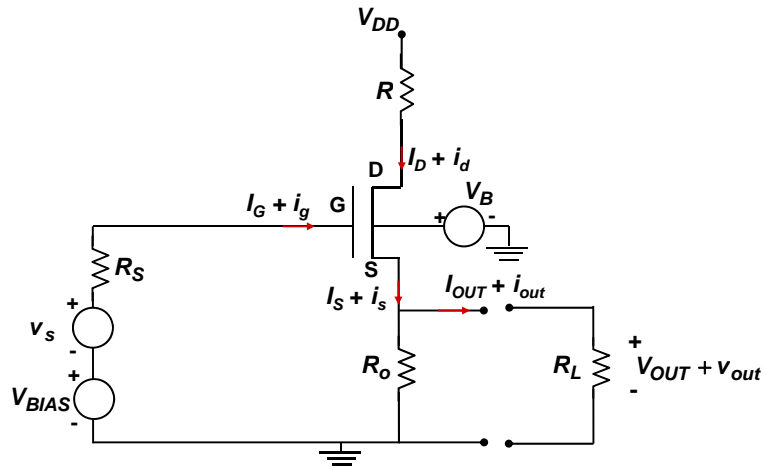
### The Common Gate Amplifier: A Transimpedance Amplifier



Compare with the standard transimpedance amplifier model:

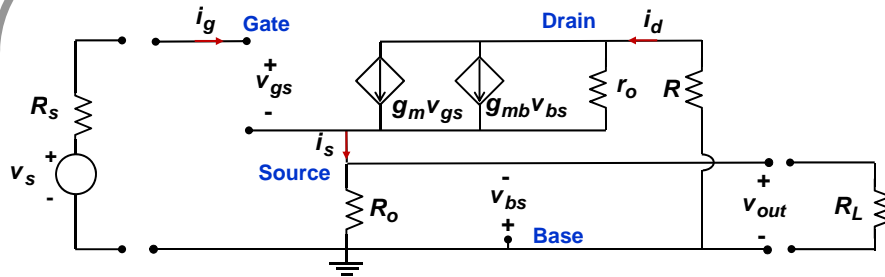


### The Common Drain Amplifier (or the Source Follower)

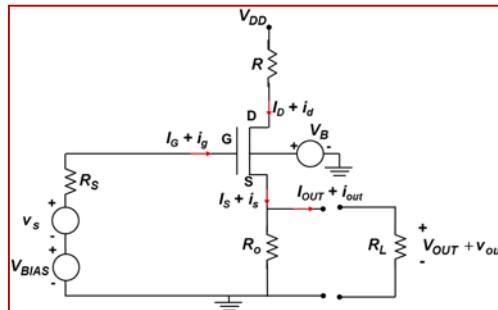


The drain terminal is "common" between the input and the output  
 The common drain amplifiers are useful when large input resistances and small output resistances are desired in voltage amplifiers  
Note: The bulk is not tied to the source

### The Common Drain Amplifier: Small Signal Model

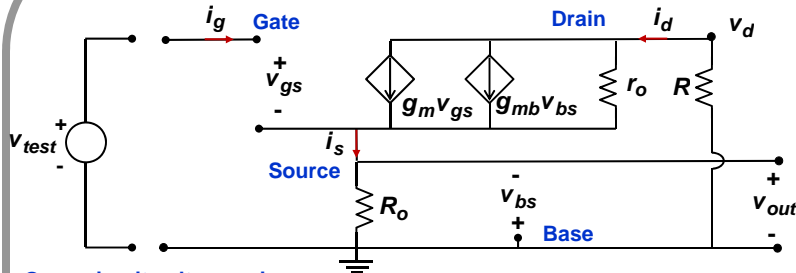


Note:  
 $v_{bs} = -v_{out}$





### The Common Drain Amplifier: Open Circuit Voltage Gain



Open circuit voltage gain:

$$v_{test} = v_{gs} + v_{out} = v_{gs} + i_d R_o$$

$$\Rightarrow \frac{v_{out}}{R_o} = i_d = \frac{(g_m r_o v_{gs} - (1 + g_{mb} r_o) v_{out})}{(r_o + R)}$$

$$\Rightarrow v_{out} \left( \frac{1}{R_o} + \frac{(1 + g_{mb} r_o)}{(r_o + R)} \right) = v_{gs} \frac{g_m r_o}{(r_o + R)} = (v_{test} - v_{out}) \frac{g_m r_o}{(r_o + R)}$$

$$\Rightarrow v_{out} \left( \frac{1}{R_o} + \frac{(1 + g_{mb} r_o)}{(r_o + R)} + \frac{g_m r_o}{(r_o + R)} \right) = v_{test} \frac{g_m r_o}{(r_o + R)}$$

$$\Rightarrow A_v = \frac{v_{out}}{v_{test}} = \frac{\frac{g_m r_o R_o}{(r_o + R)}}{1 + \frac{(g_m + g_{mb}) r_o R_o}{(r_o + R)}} < 1$$

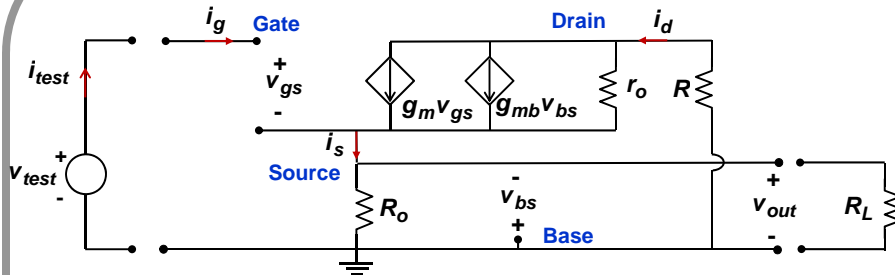
$$i_d = g_m v_{gs} - g_{mb} v_{out} + \frac{v_d - v_{out}}{r_o}$$

$$= g_m v_{gs} - g_{mb} v_{out} + \frac{-i_d R - v_{out}}{r_o}$$

$$\Rightarrow i_d = \frac{(g_m r_o v_{gs} - (1 + g_{mb} r_o) v_{out})}{(r_o + R)}$$

Less than unity – but can be very close to unity

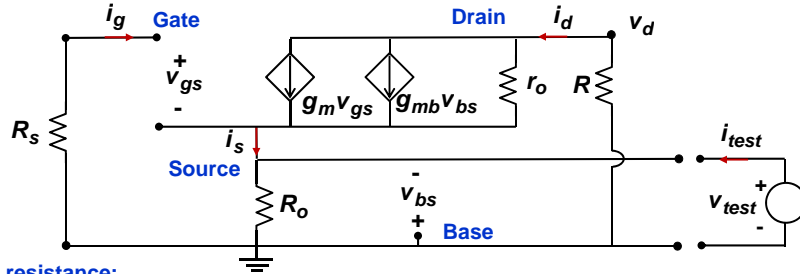
### The Common Drain Amplifier: Input Resistance



Input resistance:

$$R_{in} = \frac{v_{test}}{i_{test}} = \infty \longrightarrow \text{Large}$$

### The Common Drain Amplifier: Output Resistance



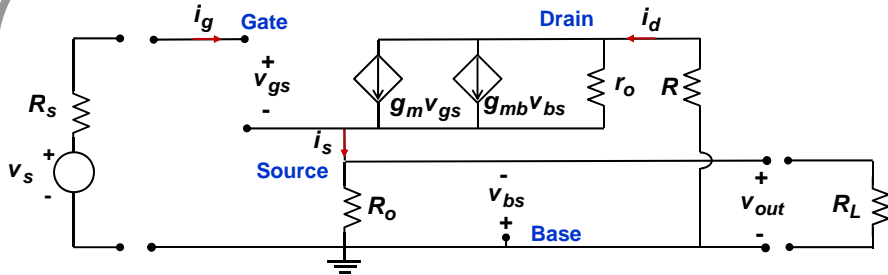
Output resistance:

$$\begin{aligned}
 i_{test} &= \frac{v_{test}}{R_o} - i_d \\
 &= \frac{v_{test}}{R_o} - \frac{(g_m r_o v_{gs} - (1 + g_{mb} r_o) v_{test})}{(r_o + R)} \\
 &= \frac{v_{test}}{R_o} + \frac{(g_m r_o + 1 + g_{mb} r_o) v_{test}}{(r_o + R)} \\
 \Rightarrow \frac{1}{R_{out}} &= \frac{i_{test}}{v_{test}} = \frac{1}{R_o} + \frac{(g_m + g_{mb}) r_o + 1}{(r_o + R)}
 \end{aligned}$$

$$\begin{aligned}
 i_d &= g_m v_{gs} - g_{mb} v_{test} + \frac{v_d - v_{test}}{r_o} \\
 &= g_m v_{gs} - g_{mb} v_{test} + \frac{-i_d R - v_{test}}{r_o} \\
 \Rightarrow i_d &= \frac{(g_m r_o v_{gs} - (1 + g_{mb} r_o) v_{test})}{(r_o + R)}
 \end{aligned}$$

$R_{out}$  can be small if:  
 $r_o \gg R$   
 $(g_m + g_{mb}) r_o \gg 1$   
 and/or if  $R_o$  is small

### The Common Drain Amplifier: A Voltage Buffer



Compare with the standard voltage amplifier model:

