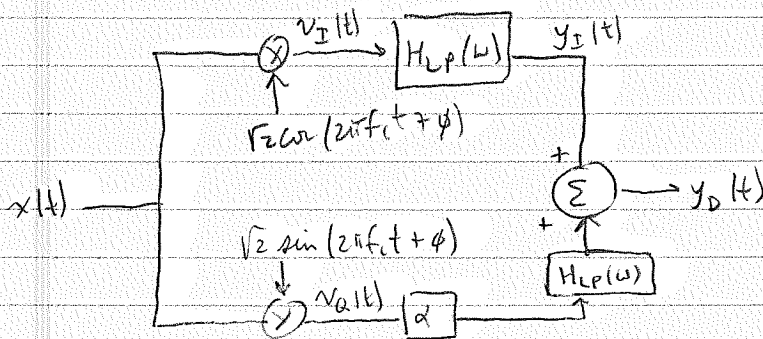
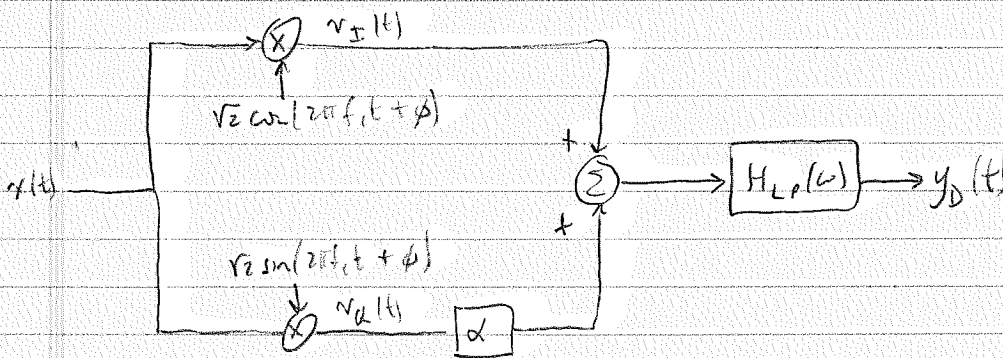


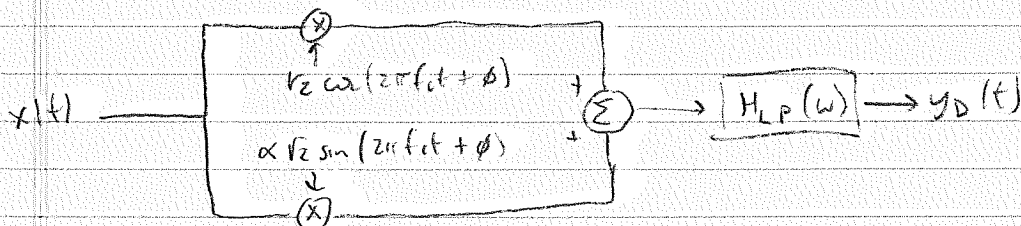
is equivalent to the following by commutative property



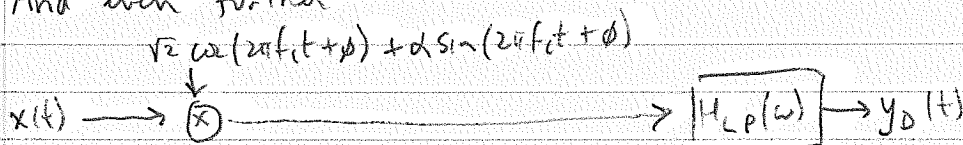
is equivalent to the following by distributive property

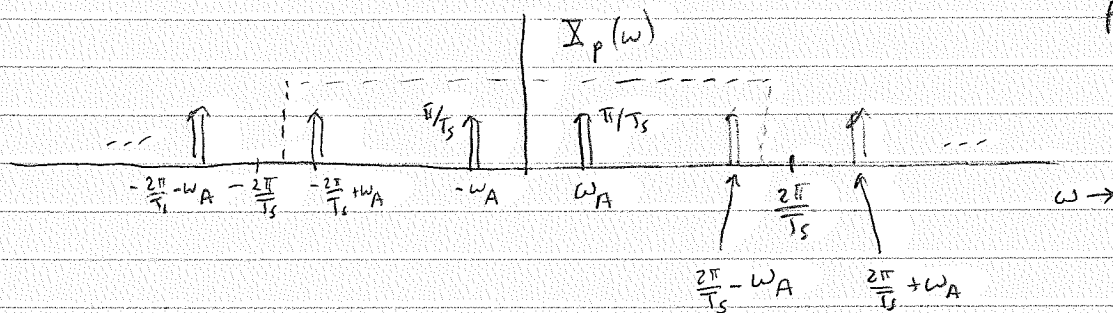


Could simplify further:



And even further





$$\omega_B = \frac{2\pi}{T_s} - \omega_A \Rightarrow \frac{2\pi}{T_s} = \omega_A + \omega_B \Rightarrow \frac{T_s}{2\pi} = \frac{1}{\omega_A + \omega_B} \Rightarrow \boxed{T_s = \frac{2\pi}{\omega_A + \omega_B}}$$

$$\frac{2\pi}{T_s} - \omega_A < \omega_H < \frac{2\pi}{T_s} + \omega_A \Rightarrow \omega_A + \omega_B - \omega_A < \omega_H < \omega_A + \omega_B + \omega_A$$

$$\Rightarrow \boxed{\omega_B < \omega_H < 2\omega_A + \omega_B}$$

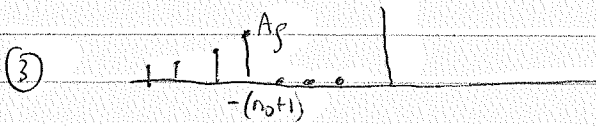
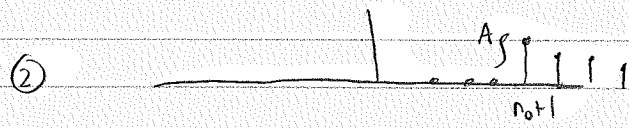
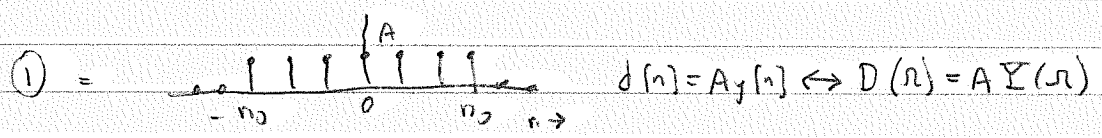
$$x[n] = \begin{cases} A & |n| \leq n_0 \\ A \rho^{|n| - n_0} & |n| > n_0 \end{cases} \quad A \in \mathbb{R}, \rho \in \mathbb{R}, |\rho| < 1$$

$$y[n] = \begin{cases} 1 & |n| \leq n_0 \\ 0 & |n| > n_0 \end{cases} \leftrightarrow \underline{Y}(\omega) = \frac{\sin(\omega(2n_0+1)/2)}{\sin(\omega/2)}$$

$$z[n] = \rho^n u[n] \leftrightarrow \underline{Z}(\omega) = \frac{1}{1 - \rho e^{-j\omega}}$$

$$(a) \quad x[n] = \underbrace{A y[n]}_{(1)} + \underbrace{A \rho^{n-n_0} u[n-n_0-1]}_{(2)} + \underbrace{A \rho^{-n-n_0} u[-n-n_0-1]}_{(3)}$$

$= 1 \text{ for } n \geq n_0+1$   
 $= 0 \text{ otherwise}$ 
 $= 1 \text{ for } n \leq -n_0-1$   
 $= 0 \text{ otherwise}$



$$\begin{aligned}
 x[n] &\leftrightarrow \underline{X}(\omega) \\
 y[n] &= x[n-n_0] \\
 \Rightarrow \underline{Y}(\omega) &= \sum_{n=-\infty}^{+\infty} y[n] e^{-j\omega n} \\
 &= \sum_{n=-\infty}^{+\infty} x[n-n_0] e^{-j\omega n} \quad m = n - n_0 \\
 &= \sum_{m=-\infty}^{+\infty} x[m] e^{-j\omega(m+n_0)} = e^{-j\omega n_0} \underline{X}(\omega)
 \end{aligned}$$

②:  $b[n] = A \rho^{n-n_0} u[n-n_0-1]$

$$c[n] = b[n+(n_0+1)] = A \rho^{n+(n_0+1)-n_0} u[n+(n_0+1)-n_0-1] = A \rho^{n+1} u[n]$$

$$= A \rho \rho^n u[n] = A \rho z[n] \leftrightarrow \underline{C}(\omega) = A \rho \underline{Z}(\omega)$$

Also,  $\underline{C}(\omega) = e^{j\omega(n_0+1)} \underline{B}(\omega)$

$$\Rightarrow \underline{B}(\omega) = e^{-j\omega(n_0+1)} \underline{C}(\omega) = e^{-j\omega(n_0+1)} A \rho \underline{Z}(\omega)$$

③:  $f[n] = b[-n]$

$$F(\omega) = \sum_{n=-\infty}^{+\infty} f[n] e^{-j\omega n} = \sum_{n=-\infty}^{+\infty} b[-n] e^{-j\omega n} \quad m = -n$$

$$= \sum_{m=-\infty}^{+\infty} b[m] e^{-j\omega(-m)} = \sum_{m=-\infty}^{+\infty} b[m] e^{-j(-\omega)m} = B(-\omega)$$

$$x[n] = d[n] + b[n] + f[n]$$

↓

$$X(\omega) = D(\omega) + B(\omega) + F(\omega)$$

$$= A \Psi(\omega) + B(\omega) + B(-\omega)$$

$b[n] \in \mathbb{R}$

$$B(\omega) = \sum_{n=-\infty}^{+\infty} b[n] e^{-j\omega n}$$

$$\Rightarrow B^*(\omega) = \sum_{n=-\infty}^{+\infty} b^*[n] e^{+j\omega n} = \sum_{n=-\infty}^{+\infty} b[n] e^{-j(-\omega)n} = B(-\omega)$$

Alternatively, look at the formulas for  $B(\omega)$  and  $Z(\omega)$  and see that  $B^*(\omega) = B(-\omega)$ .

Either way,

$$X(\omega) = A \Psi(\omega) + B(\omega) + B^*(\omega)$$

$$= A \Psi(\omega) + 2 \operatorname{Re}\{B(\omega)\}$$

$$= A \Psi(\omega) + 2 \operatorname{Re}\left\{ \exp(-j\omega(n_0+1)) A_g Z(\omega) \right\}$$

$$= A \Psi(\omega) + 2 A_g \operatorname{Re}\left\{ \exp(-j\omega(n_0+1)) \frac{1}{1 - pe^{-j\omega}} \right\}$$

(b) In the calculations of (a) it is shown that

$$x[n] \in \mathbb{R} \Rightarrow \bar{X}^*(\Omega) = \bar{X}(-\Omega).$$

$$x[n] = x[-n] \text{ (i.e., } x[n] \text{ is even)} \Rightarrow \bar{X}(\Omega) = \bar{X}(-\Omega).$$

Therefore,

$$\bar{X}^*(\Omega) = \bar{X}(-\Omega) = \bar{X}(\Omega).$$

Therefore

$$\bar{X}(\Omega) \in \mathbb{R}.$$

(c) So it is necessary to show that  $\bar{X}(\Omega)$  from (a) is real.

This is straightforward since

$$\bar{X}(\Omega) = A \bar{Y}(\Omega) + 2A \rho \operatorname{Re} \left\{ \exp(-j\Omega(n_0+1)) \frac{1}{1-\rho e^{-j\Omega}} \right\}$$

$\uparrow \quad \uparrow \quad \uparrow \quad \uparrow \quad \uparrow$   
 real real real real real

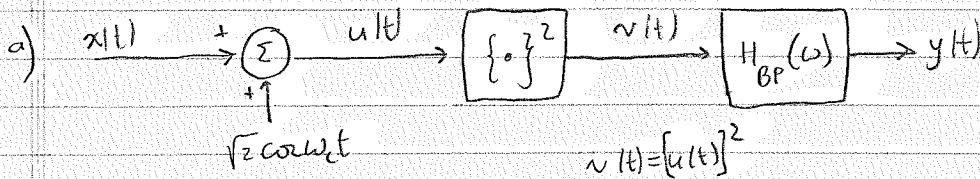
So  $\bar{X}(\Omega) \in \mathbb{R}$



$$\cos^2 \alpha = \frac{1}{2} (1 + \cos 2\alpha) \quad \cos \alpha \sin \beta = \frac{1}{2} \sin(\alpha + \beta) - \frac{1}{2} \sin(\alpha - \beta)$$

$$\sin^2 \alpha = \frac{1}{2} (1 - \cos 2\alpha)$$

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$$x(t) = A_c m(t)$$

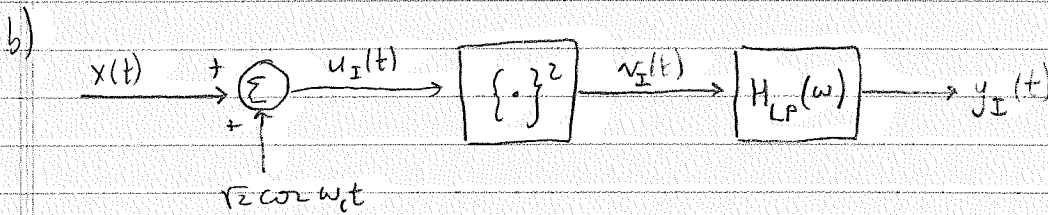
$$u(t) = A_c m(t) + \sqrt{2} \cos \omega_c t$$

$$v(t) = A_c^2 m^2(t) + 2A_c m(t) \sqrt{2} \cos \omega_c t + 2 \cos^2 \omega_c t$$

$$= A_c^2 m^2(t) + 2A_c m(t) \sqrt{2} \cos \omega_c t + 1 + \cos 2\omega_c t$$

$$= \underbrace{1 + A_c^2 m^2(t)}_{\text{below filter's cutoff}} + \underbrace{2A_c m(t) \sqrt{2} \cos \omega_c t}_{\text{in filter's passband}} + \underbrace{\cos 2\omega_c t}_{\text{above filter's cutoff}}$$

$$y(t) = 2A_c m(t) \sqrt{2} \cos \omega_c t$$



$$x(t) = A_c m(t) \sqrt{2} \cos \omega_c t$$

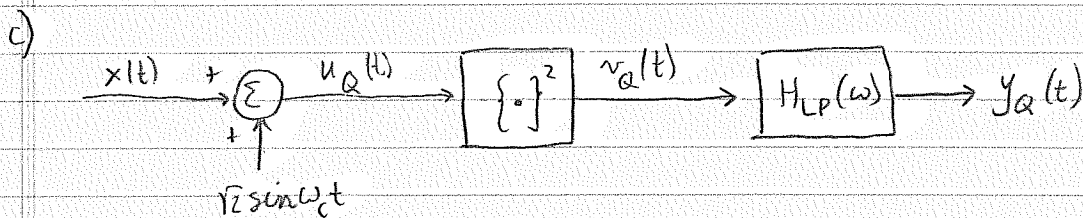
$$u_I(t) = (A_c m(t) + 1) \sqrt{2} \cos \omega_c t$$

$$v_I(t) = (A_c^2 m^2(t) + 2A_c m(t) + 1) 2 \cos^2 \omega_c t$$

$$= (A_c^2 m^2(t) + 2A_c m(t) + 1) (1 + \cos 2\omega_c t)$$

$$= \underbrace{(A_c^2 m^2(t) + 2A_c m(t) + 1)}_{\text{in filter's passband}} + \underbrace{(A_c^2 m^2(t) + 2A_c m(t) + 1) \cos(2\omega_c t)}_{\text{in filter's stop band}}$$

$$y_I(t) = \underbrace{A_c^2 m^2(t) + 1}_{\text{junk!}} + \underbrace{2A_c m(t)}_{\text{desired signal}} = [A_c m(t) + 1]^2$$



$$x(t) = A_c m(t) \sqrt{2} \cos \omega_c t$$

$$u_Q(t) = A_c m(t) \sqrt{2} \cos \omega_c t + \sqrt{2} \sin \omega_c t$$

$$\begin{aligned}
 r_Q(t) &= A_c^2 m^2(t) 2 \cos^2 \omega_c t + 2 A_c m(t) \sqrt{2} \cos \omega_c t \sqrt{2} \sin \omega_c t + 2 \sin^2 \omega_c t \\
 &= A_c^2 m^2(t) (1 + \cos 2\omega_c t) + 4 A_c m(t) \left[ \frac{1}{2} \sin(2\omega_c t) - \frac{1}{2} \underbrace{\sin 0}_{=0} \right] \\
 &\quad + [1 - \cos 2\omega_c t] \\
 &= A_c^2 m^2(t) + 1 + A_c^2 m^2(t) \cos 2\omega_c t + 2 A_c m(t) \sin 2\omega_c t - \cos 2\omega_c t \\
 &= \underbrace{A_c^2 m^2(t) + 1}_{\text{in filter's passband}} + \underbrace{(A_c^2 m^2(t) - 1) \cos 2\omega_c t + 2 A_c m(t) \sin 2\omega_c t}_{\text{in filter's stopband}}
 \end{aligned}$$

$$y_Q(t) = A_c^2 m^2(t) + 1$$

(d) Note that  $y(t)$  defined by  $y(t) = y_I(t) - y_Q(t)$  has the value

$$\begin{aligned}
 y(t) &= y_I(t) - y_Q(t) \\
 &= A_c^2 m^2(t) + 1 + 2 A_c m(t) - [A_c^2 m^2(t) + 1] \\
 &= 2 A_c m(t)
 \end{aligned}$$

which is the signal that we want.

The corresponding block diagram is

