

Lecture 25

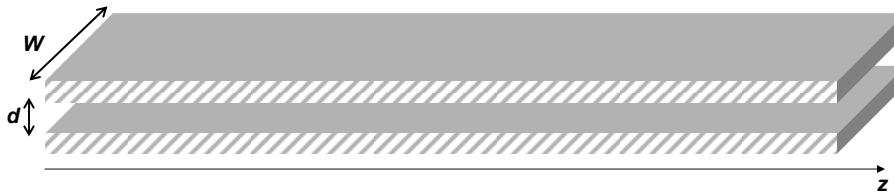
Guided Waves in Parallel Plate Metal Waveguides

In this lecture you will learn:

- Parallel plate metal waveguides
- TE and TM guided modes in waveguides

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Parallel Plate Metal Waveguides

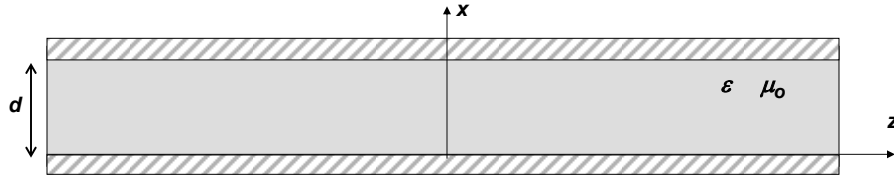


- Consider a parallel plate waveguide (shown above)
- We have studied such structures in the context of [transmission lines](#)
- We know that they can guide **TEM** waves (**T**ransverse **E**lectric and **M**agnetic) in which both the electric and magnetic fields point in direction perpendicular to the propagation direction
- But these structures can guide more than just the **TEM** waves that we have considered so far

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Basic Wave Equations

Consider a parallel plate waveguide:



The electric field of any guided wave will satisfy the complex wave equations:

$$\left. \begin{aligned} \nabla \times \vec{E}(\vec{r}) &= -j\omega \mu_0 \vec{H}(\vec{r}) \\ \nabla \times \vec{H}(\vec{r}) &= j\omega \epsilon \vec{E}(\vec{r}) \end{aligned} \right\} \longrightarrow \begin{aligned} \nabla^2 \vec{E}(\vec{r}) &= -\omega^2 \mu_0 \epsilon \vec{E}(\vec{r}) \\ \nabla^2 \vec{H}(\vec{r}) &= -\omega^2 \mu_0 \epsilon \vec{H}(\vec{r}) \end{aligned}$$

We look for solutions of the equation,

$$\nabla^2 \vec{E}(\vec{r}) = -\omega^2 \mu_0 \epsilon \vec{E}(\vec{r})$$

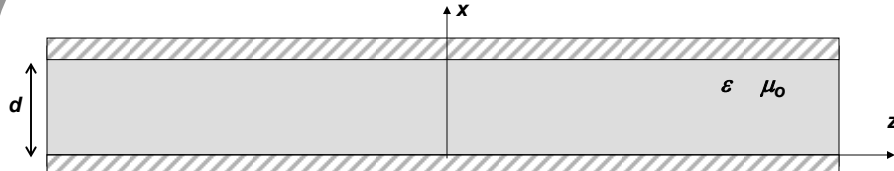
where the z-dependence is that of a wave going in the z-direction, and where the E-field is pointing in the y-direction:

$$\vec{E}(\vec{r}) = \hat{y} F(x) e^{-jk_z z}$$

→ Some unknown function of "x"

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TE Guided Modes - I



The assumed solution form: $\vec{E}(\vec{r}) = \hat{y} F(x) e^{-jk_z z}$

represents a **TE** guided wave (**T**ransverse **E**lectric) since the direction of E-field is transverse to the direction of wave propagation

Plugging the assumed solution into the equation gives:

$$\begin{aligned} \nabla^2 \vec{E}(\vec{r}) &= -\omega^2 \mu_0 \epsilon \vec{E}(\vec{r}) \\ \Rightarrow \left(\frac{\partial^2}{\partial z^2} + \frac{\partial^2}{\partial x^2} \right) \vec{E}(\vec{r}) &= -\omega^2 \mu_0 \epsilon \vec{E}(\vec{r}) \\ \Rightarrow -\frac{\partial^2 F(x)}{\partial x^2} &= (\omega^2 \mu_0 \epsilon - k_z^2) F(x) \end{aligned}$$

Perfect metal boundary conditions $\Rightarrow F(x=0) = F(x=d) = 0$

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TE Guided Modes - II

Need to solve: $-\frac{\partial^2 F(x)}{\partial x^2} = (\omega^2 \mu_0 \epsilon - k_z^2) F(x)$

With boundary conditions $\Rightarrow F(x=0) = F(x=d) = 0$

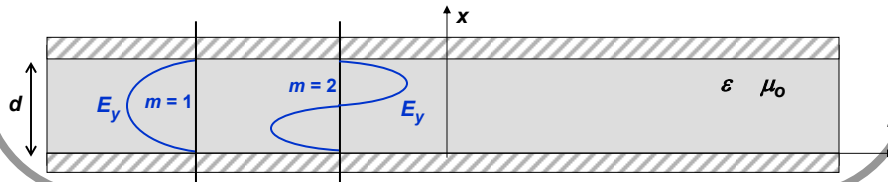
Solution is: $F(x) = E_0 \sin(k_x x)$ } Automatically satisfies the boundary condition: $F(x=0) = 0$

But the value of k_x cannot be arbitrary – boundary condition at $x = d$ dictates that:

$k_x = \frac{m\pi}{d}$ where: $m = 1, 2, 3, \dots$

Solution becomes: $F(x) = E_0 \sin\left(\frac{m\pi}{d} x\right)$

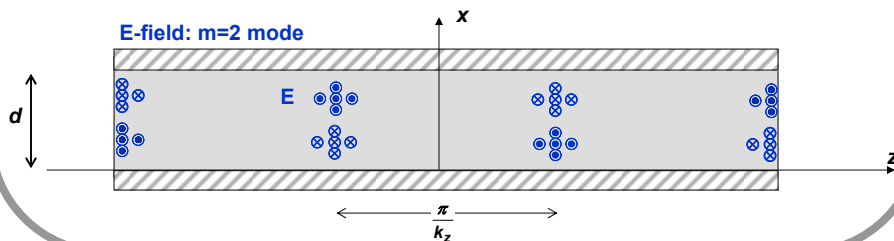
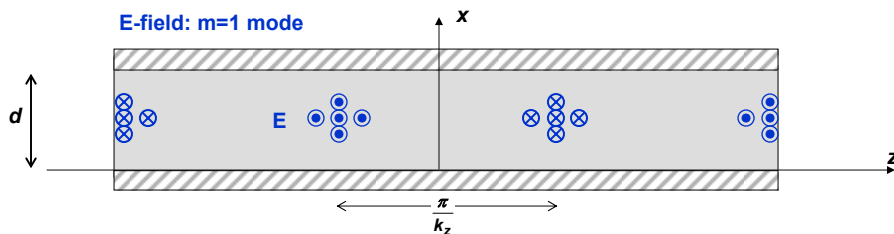
And: $\vec{E}(\vec{r}) = \hat{y} E_0 \sin\left(\frac{m\pi}{d} x\right) e^{-jk_z z}$ { $m = 1, 2, 3, \dots$



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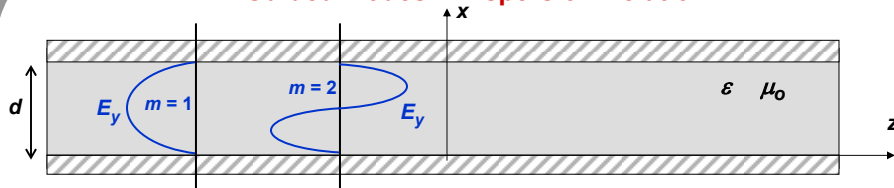
TE Guided Modes - III

$\vec{E}(\vec{r}) = \hat{y} E_0 \sin\left(\frac{m\pi}{d} x\right) e^{-jk_z z}$ { $m = 1, 2, 3, \dots$



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TE Guided Modes – Dispersion Relation



$$\vec{E}(\vec{r}) = \hat{y} E_0 \sin\left(\frac{m\pi}{d}x\right) e^{-jk_z z} \quad \{ m = 1, 2, 3, \dots \}$$

Different “m” values correspond to different TE modes – labeled as TE_m modes

The equation: $-\frac{\partial^2 F(x)}{\partial x^2} = (\omega^2 \mu_0 \varepsilon - k_z^2)F(x)$ implies:

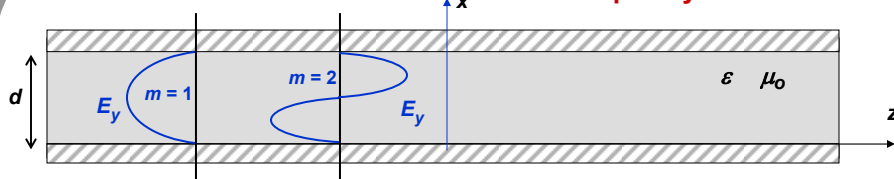
$$k_z^2 + k_x^2 = \omega^2 \mu_0 \varepsilon$$

$$\Rightarrow k_z^2 + \left(\frac{m\pi}{d}\right)^2 = \omega^2 \mu_0 \varepsilon$$

$$\Rightarrow k_z = \sqrt{\omega^2 \mu_0 \varepsilon - \left(\frac{m\pi}{d}\right)^2} \quad \left. \vphantom{\omega^2 \mu_0 \varepsilon} \right\} \text{Dispersion relation for TE}_m \text{ guided mode}$$

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TE Guided Modes – Cut-off Frequency



$$k_z = \sqrt{\omega^2 \mu_0 \varepsilon - \left(\frac{m\pi}{d}\right)^2} \quad \left. \vphantom{\omega^2 \mu_0 \varepsilon} \right\} \text{Dispersion relation for TE}_m \text{ guided mode}$$

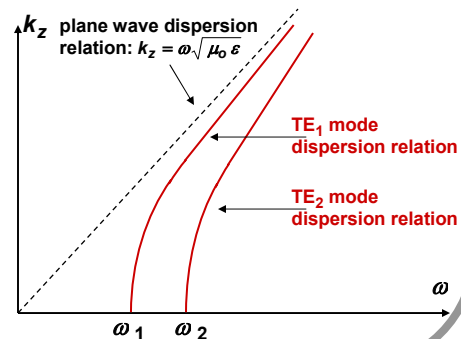
For the TE_m mode, if the frequency ω is less than:

$$\frac{1}{\sqrt{\mu_0 \varepsilon}} \left(\frac{m\pi}{d}\right)$$

Then k_z becomes entirely imaginary and the mode does not propagate (but decays exponentially with distance)

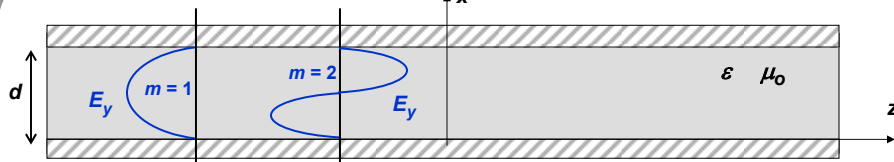
\Rightarrow Cut-off frequency for TE_m mode:

$$\omega_m = \frac{1}{\sqrt{\mu_0 \varepsilon}} \left(\frac{m\pi}{d}\right)$$



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TE Guided Modes – Magnetic Field



$$\vec{E}(\vec{r}) = \hat{y} E_0 \sin\left(\frac{m\pi}{d} x\right) e^{-jk_z z} \quad \{ m = 1, 2, 3, \dots \}$$

Magnetic field is given by the equation: $\nabla \times \vec{E}(\vec{r}) = -j\omega \mu_0 \vec{H}(\vec{r})$

$$\vec{H}(\vec{r}) = \frac{jE_0}{\omega \mu_0} \left[\hat{z} \frac{m\pi}{d} \cos\left(\frac{m\pi}{d} x\right) + \hat{x} j k_z \sin\left(\frac{m\pi}{d} x\right) \right] e^{-jk_z z}$$

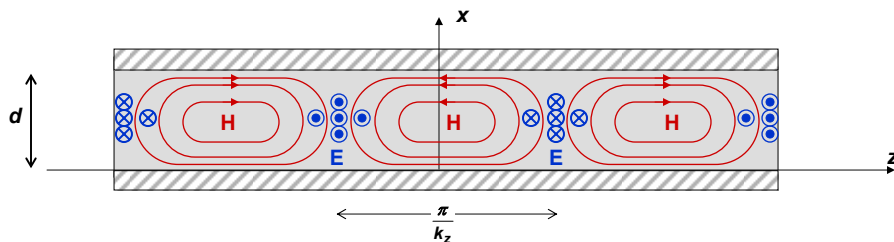
Note that the perfect metal boundary condition for the magnetic field is automatically satisfied i.e:

$$H_x(\vec{r})|_{x=0} = H_x(\vec{r})|_{x=d} = 0$$

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TE Guided Modes – Field Profiles

The E-field and H-field lines for the TE₁ mode are shown below:



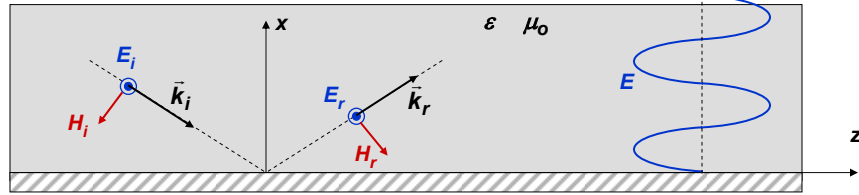
$$\vec{E}(\vec{r}) = \hat{y} E_0 \sin\left(\frac{m\pi}{d} x\right) e^{-jk_z z} \quad \{ m = 1, 2, 3, \dots \}$$

$$\vec{H}(\vec{r}) = \frac{jE_0}{\omega \mu_0} \left[\hat{z} \frac{m\pi}{d} \cos\left(\frac{m\pi}{d} x\right) + \hat{x} j k_z \sin\left(\frac{m\pi}{d} x\right) \right] e^{-jk_z z}$$

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TE Guided Modes – Another Perspective - I

Consider TE-wave reflection off a perfect metal:



$$\vec{k}_i = -k_x \hat{x} + k_z \hat{z} \quad \vec{k}_r = k_x \hat{x} + k_z \hat{z}$$

$$k_z^2 + k_x^2 = \omega^2 \mu_0 \epsilon$$

$$\vec{E}(\vec{r})|_{x>0} = \hat{y} E_i e^{-j(-k_x x + k_z z)} + \hat{y} \Gamma E_i e^{-j(k_x x + k_z z)} \longrightarrow \left\{ \Gamma = -1 \right.$$

$$\vec{E}(\vec{r})|_{x>0} = \hat{y} E_i \left[e^{-j(-k_x x + k_z z)} - e^{-j(k_x x + k_z z)} \right]$$

$$\Rightarrow \vec{E}(\vec{r})|_{x>0} = \hat{y} 2j E_i \sin(k_x x) e^{-j k_z z}$$

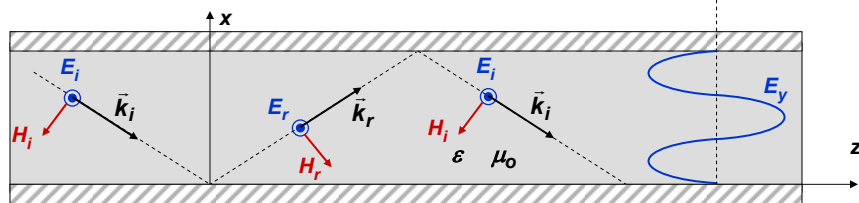
Notice the "sine" variation of the y-component of the E-field

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TE Guided Modes – Another Perspective - II

If another top metal plate is placed at one of the nodes of the "sine" function then this additional metal plate will not disturb the field

$$\vec{E}(\vec{r})|_{x>0} = \hat{y} 2j E_i \sin(k_x x) e^{-j k_z z}$$



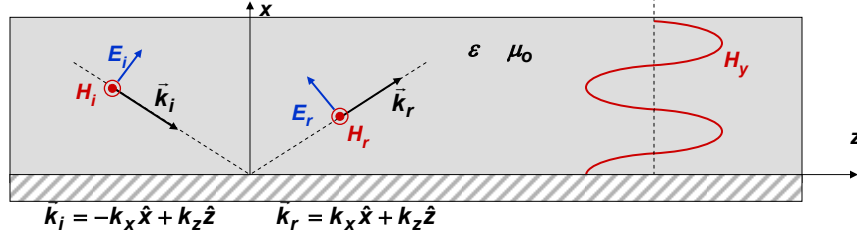
$$\vec{k}_i = -k_x \hat{x} + k_z \hat{z} \quad \vec{k}_r = k_x \hat{x} + k_z \hat{z}$$

This is exactly what guided TE modes are – TE-waves bouncing back and fourth between two metal plates and propagating in the z-direction !

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TM Guided Modes - I

Consider TM-wave reflection off a perfect metal:



$$k_z^2 + k_x^2 = \omega^2 \mu_0 \epsilon$$

$$\bar{H}(\vec{r})|_{x>0} = \hat{y} H_i e^{-j(-k_x x + k_z z)} + \hat{y} \Gamma^{TM} H_i e^{-j(k_x x + k_z z)} \longrightarrow \left\{ \Gamma^{TM} = +1 \right.$$

$$\bar{H}(\vec{r})|_{x>0} = \hat{y} H_i \left[e^{-j(-k_x x + k_z z)} + e^{-j(k_x x + k_z z)} \right]$$

$$\Rightarrow \bar{H}(\vec{r})|_{x>0} = \hat{y} 2H_i \cos(k_x x) e^{-j k_z z}$$

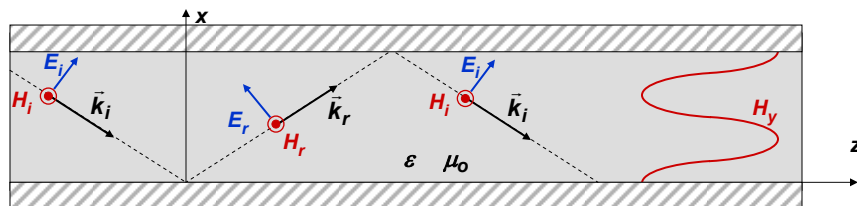
Notice the "cosine" variation of the y-component of the H-field

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TM Guided Modes - II

If another top metal plate is placed at the maximum points of the "cosine" function then this additional metal plate will not disturb the field

$$\bar{H}(\vec{r})|_{x>0} = \hat{y} 2H_i \cos(k_x x) e^{-j k_z z}$$

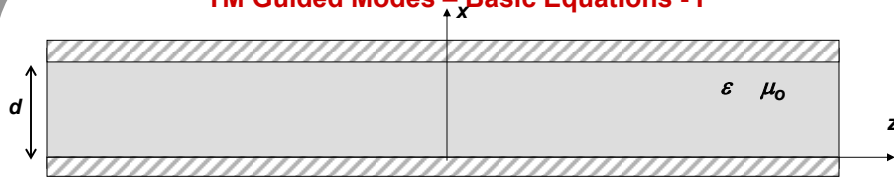


$$\bar{k}_i = -k_x \hat{x} + k_z \hat{z} \quad \bar{k}_r = k_x \hat{x} + k_z \hat{z}$$

This is exactly what guided TM modes are – TM-waves bouncing back and fourth between two metal plates and propagating in the z-direction !

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TM Guided Modes – Basic Equations - I



Need to solve the equation: $\nabla^2 \vec{H}(\vec{r}) = -\omega^2 \mu_0 \epsilon \vec{H}(\vec{r})$

Assume the solution form: $\vec{H}(\vec{r}) = \hat{y} G(x) e^{-j k_z z}$

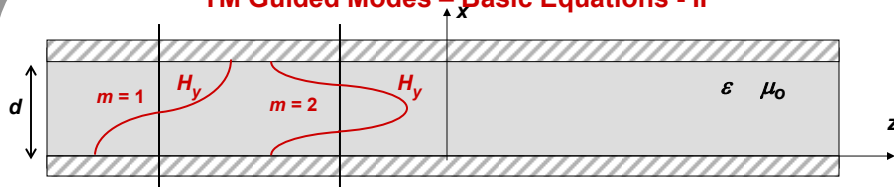
It represents a **TM** guided wave (**T**ransverse **M**agnetic) since the direction of H-field is transverse to the direction of wave propagation

Plugging the assumed solution into the equation gives:

$$\begin{aligned} \nabla^2 \vec{H}(\vec{r}) &= -\omega^2 \mu_0 \epsilon \vec{H}(\vec{r}) \\ \Rightarrow \left(\frac{\partial^2}{\partial z^2} + \frac{\partial^2}{\partial x^2} \right) \vec{H}(\vec{r}) &= -\omega^2 \mu_0 \epsilon \vec{H}(\vec{r}) \\ \Rightarrow -\frac{\partial^2 G(x)}{\partial x^2} &= (\omega^2 \mu_0 \epsilon - k_z^2) G(x) \end{aligned}$$

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TM Guided Modes – Basic Equations - II



Need to solve: $-\frac{\partial^2 G(x)}{\partial x^2} = (\omega^2 \mu_0 \epsilon - k_z^2) G(x)$

Solution is: $G(x) = H_0 \cos(k_x x)$ } Motivation for this is obtained from the TM-wave reflection analysis discussed earlier

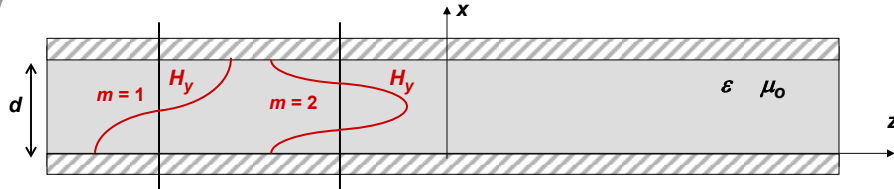
$$k_x = m \frac{\pi}{d} \quad \text{where : } m = 0, 1, 2, 3, \dots$$

Solution becomes: $G(x) = H_0 \cos\left(\frac{m \pi}{d} x\right)$

And: $\vec{H}(\vec{r}) = \hat{y} H_0 \cos\left(\frac{m \pi}{d} x\right) e^{-j k_z z}$ { $m = 0, 1, 2, 3, \dots$

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TM Guided Modes – Electric Field



$$\vec{H}(\vec{r}) = \hat{y} H_0 \cos\left(\frac{m\pi}{d} x\right) e^{-jk_z z} \quad \{ m = 0, 1, 2, 3, \dots \}$$

Electric field is given by the equation: $\nabla \times \vec{H}(\vec{r}) = j\omega \epsilon \vec{E}(\vec{r})$

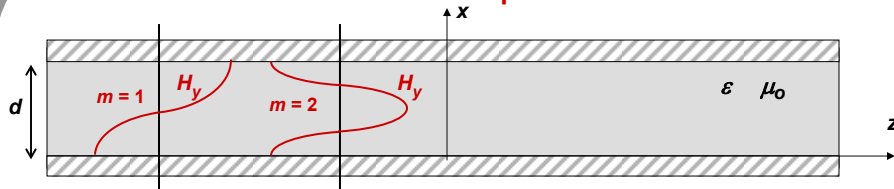
$$\vec{E}(\vec{r}) = -\frac{jH_0}{\omega \epsilon} \left[-\hat{z} \frac{m\pi}{d} \sin\left(\frac{m\pi}{d} x\right) + \hat{x} j k_z \cos\left(\frac{m\pi}{d} x\right) \right] e^{-jk_z z}$$

Note that the perfect metal boundary condition for the electric field is automatically satisfied, i.e.:

$$E_z(\vec{r})_{x=0} = E_z(\vec{r})_{x=d} = 0$$

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TM Guided Modes – Dispersion Relation



$$\vec{H}(\vec{r}) = \hat{y} H_0 \cos\left(\frac{m\pi}{d} x\right) e^{-jk_z z} \quad \{ m = 0, 1, 2, 3, \dots \}$$

Different “m” values correspond to different TM modes – labeled as TM_m modes

The equation: $-\frac{\partial^2 G(x)}{\partial x^2} = (\omega^2 \mu_0 \epsilon - k_z^2) G(x)$ implies:

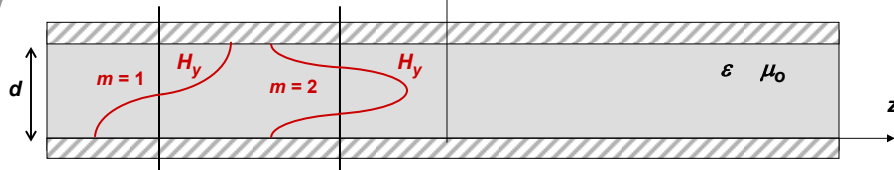
$$k_z^2 + k_x^2 = \omega^2 \mu_0 \epsilon$$

$$\Rightarrow k_z^2 + \left(\frac{m\pi}{d}\right)^2 = \omega^2 \mu_0 \epsilon$$

$$\Rightarrow k_z = \sqrt{\omega^2 \mu_0 \epsilon - \left(\frac{m\pi}{d}\right)^2} \quad \left. \vphantom{\Rightarrow} \right\} \text{Dispersion relation for } TM_m \text{ waveguide mode}$$

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TM Guided Modes – Cut-off Frequency



$$k_z = \sqrt{\omega^2 \mu_0 \varepsilon - \left(\frac{m \pi}{d}\right)^2} \quad \left. \vphantom{k_z} \right\} \text{Dispersion relation for } \text{TM}_m \text{ guided mode}$$

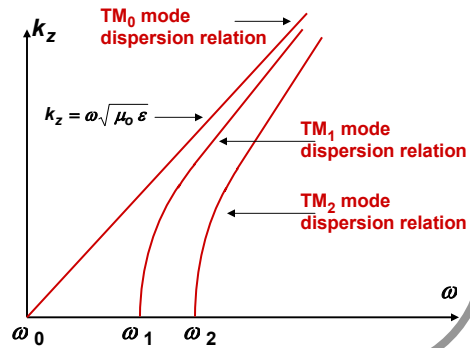
For the TM_m mode, if the frequency ω is less than:

$$\frac{1}{\sqrt{\mu_0 \varepsilon}} \left(\frac{m \pi}{d}\right)$$

Then k_z becomes entirely imaginary and the mode does not propagate (but decays exponentially with distance)

⇒ **Cut-off frequency** for TM_m mode:

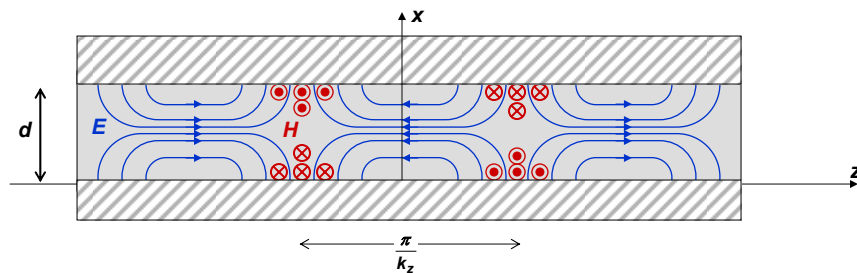
$$\omega_m = \frac{1}{\sqrt{\mu_0 \varepsilon}} \left(\frac{m \pi}{d}\right)$$



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TM Guided Modes – Field Profiles

The **E-field** and **H-field** lines for the TM_1 mode are shown below:



$$\vec{H}(\vec{r}) = \hat{y} H_0 \cos\left(\frac{m \pi}{d} x\right) e^{-j k_z z} \quad \{ m = 0, 1, 2, 3, \dots \}$$

$$\vec{E}(\vec{r}) = -\frac{j H_0}{\omega \varepsilon} \left[-\hat{z} \frac{m \pi}{d} \sin\left(\frac{m \pi}{d} x\right) + \hat{x} j k_z \cos\left(\frac{m \pi}{d} x\right) \right] e^{-j k_z z}$$

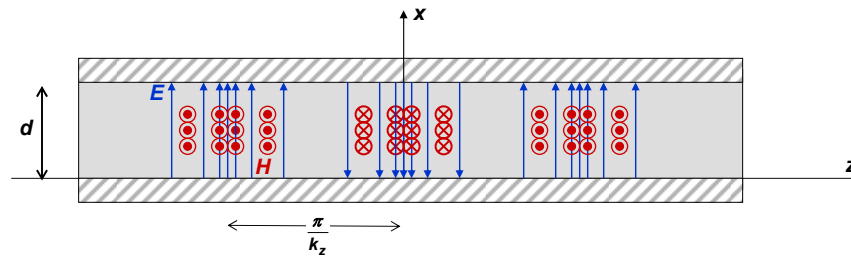
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TM₀ Guided Mode – Field Profiles

The E-field and H-field for the TM₀ mode are:

$$\begin{aligned}\vec{H}(\vec{r}) &= \hat{y} H_0 e^{-j k_z z} \\ \vec{E}(\vec{r}) &= \hat{x} \frac{k_z}{\omega \epsilon} H_0 e^{-j k_z z}\end{aligned} \quad \left. \vphantom{\begin{aligned}\vec{H}(\vec{r}) \\ \vec{E}(\vec{r})\end{aligned}} \right\} \text{Note that fields are not a function of "x"}$$

The E-field and H-field lines for the TM₀ mode are shown below:



The TM₀ mode is just the TEM mode that we worked with when dealing with transmission lines !