

Lecture 16

Waves in Isotropic Media: Dielectrics and Conductors

In this lecture you will learn:

- Wave propagation in dielectric media
- Waves propagation in conductive media

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Review: Plane Waves in Free Space

Faraday's Law:

$$\nabla \times \vec{E}(\vec{r}) = -j \omega \mu_0 \vec{H}(\vec{r})$$

Ampere's Law:

$$\nabla \times \vec{H}(\vec{r}) = \vec{J}(\vec{r}) + j \omega \epsilon_0 \vec{E}(\vec{r})$$

Complex Wave Equation: Assume: $\vec{J}(\vec{r}) = \rho(\vec{r}) = 0$

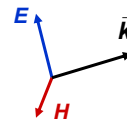
$$\nabla \times \nabla \times \vec{E}(\vec{r}) = -j \omega \mu_0 \nabla \times \vec{H}(\vec{r}) = \omega^2 \mu_0 \epsilon_0 \vec{E}(\vec{r})$$

$$\Rightarrow \nabla(\nabla \cdot \vec{E}(\vec{r})) - \nabla^2 \vec{E}(\vec{r}) = \omega^2 \mu_0 \epsilon_0 \vec{E}(\vec{r})$$

$$\Rightarrow \nabla^2 \vec{E}(\vec{r}) = -\omega^2 \mu_0 \epsilon_0 \vec{E}(\vec{r})$$

For a plane wave in free space we know the E-field and H-field phasors to be:

$$\left. \begin{aligned} \vec{E}(\vec{r}) &= \hat{n} E_0 e^{-j \vec{k} \cdot \vec{r}} \\ \vec{H}(\vec{r}) &= (\hat{k} \times \hat{n}) \frac{E_0}{\eta_0} e^{-j \vec{k} \cdot \vec{r}} \end{aligned} \right\} \begin{aligned} k &= \omega \sqrt{\mu_0 \epsilon_0} = \frac{\omega}{c} \\ \eta_0 &= \sqrt{\frac{\mu_0}{\epsilon_0}} \approx 377 \Omega \end{aligned}$$



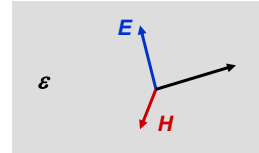
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Waves in a Dielectric Medium – Wave Equation

Suppose we have a plane wave of the form,

$$\vec{E}(\vec{r}) = \hat{n} E_0 e^{-j\vec{k}\cdot\vec{r}}$$

traveling in an infinite dielectric medium with permittivity ϵ



What is different from wave propagation in free space?

Faraday's Law:

$$\nabla \times \vec{E}(\vec{r}) = -j\omega \mu_0 \vec{H}(\vec{r})$$

Ampere's Law:

$$\nabla \times \vec{H}(\vec{r}) = \vec{J}(\vec{r}) + j\omega \epsilon \vec{E}(\vec{r})$$

Complex Wave Equation: Assume: $\vec{J}(\vec{r}) = \rho(\vec{r}) = 0$

$$\begin{aligned} \nabla \times \nabla \times \vec{E}(\vec{r}) &= -j\omega \mu_0 \nabla \times \vec{H}(\vec{r}) = \omega^2 \mu_0 \epsilon \vec{E}(\vec{r}) \\ \Rightarrow \nabla(\nabla \cdot \vec{E}(\vec{r})) - \nabla^2 \vec{E}(\vec{r}) &= \omega^2 \mu_0 \epsilon \vec{E}(\vec{r}) \\ \Rightarrow \nabla^2 \vec{E}(\vec{r}) &= -\omega^2 \mu_0 \epsilon \vec{E}(\vec{r}) \end{aligned} \quad \left\{ \begin{array}{l} \text{compare with the} \\ \text{complex wave equation} \\ \text{in free space} \\ \nabla^2 \vec{E}(\vec{r}) = -\omega^2 \mu_0 \epsilon_0 \vec{E}(\vec{r}) \end{array} \right.$$

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Waves in a Dielectric Medium – Dispersion Relation

Substitute the plane wave solution:

$$\vec{E}(\vec{r}) = \hat{n} E_0 e^{-j\vec{k}\cdot\vec{r}}$$

in the complex wave equation:

$$\nabla^2 \vec{E}(\vec{r}) = -\omega^2 \mu_0 \epsilon \vec{E}(\vec{r})$$

To get:

$$\begin{aligned} \nabla^2 \vec{E}(\vec{r}) &= -\omega^2 \mu_0 \epsilon \vec{E}(\vec{r}) \\ \Rightarrow -\vec{k} \cdot \vec{k} \vec{E}(\vec{r}) &= -\omega^2 \mu_0 \epsilon \vec{E}(\vec{r}) \\ \Rightarrow k^2 &= \omega^2 \mu_0 \epsilon \end{aligned}$$

$$\Rightarrow k = \omega \sqrt{\mu_0 \epsilon} \quad \left\{ \begin{array}{l} \text{compare with } k = \omega \sqrt{\mu_0 \epsilon_0} \\ \text{for waves in free space} \end{array} \right.$$

Refractive Index:

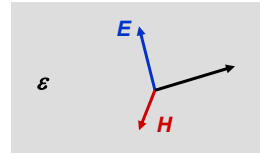
Define refractive index “n” of a dielectric medium as: $n = \sqrt{\frac{\epsilon}{\epsilon_0}}$

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Waves in a Dielectric Medium – Velocity

Plane wave:

$$\vec{E}(\vec{r}) = \hat{n} E_0 e^{-j\vec{k}\cdot\vec{r}}$$



Dispersion relation:

$$k = \omega \sqrt{\mu_0 \epsilon}$$

$$\Rightarrow k = \omega \sqrt{\mu_0 \epsilon_0} \sqrt{\frac{\epsilon}{\epsilon_0}} \longrightarrow \left\{ n = \sqrt{\frac{\epsilon}{\epsilon_0}} \right.$$

$$\Rightarrow k = \omega \frac{n}{c} \longrightarrow \left\{ \text{compare with } k = \frac{\omega}{c} \text{ for waves in free space} \right.$$

The velocity of waves in a dielectric medium is reduced from the velocity of waves in free space by the refractive index

• Velocity of waves in free space: c

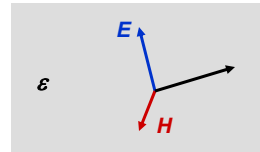
• Velocity of waves in dielectric medium of refractive index n : $\frac{c}{n}$

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Waves in a Dielectric Medium - Wavelength

Plane wave in a dielectric medium:

$$\vec{E}(\vec{r}) = \hat{n} E_0 e^{-j\vec{k}\cdot\vec{r}}$$



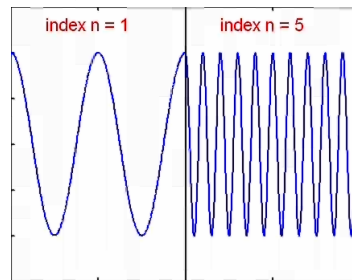
Dispersion relation: $k = \omega \frac{n}{c}$

But the magnitude of the wavevector is related to the wavelength by the relation: $k = \frac{2\pi}{\lambda}$

So for a dielectric medium we get:

$$\lambda = \frac{2\pi c}{\omega n} \longrightarrow \left\{ \text{compare with } \lambda = \frac{2\pi c}{\omega} \text{ for waves in free space} \right.$$

The wavelength of plane waves in a dielectric medium is reduced from the wavelength of plane waves of the same frequency in free space by the refractive index

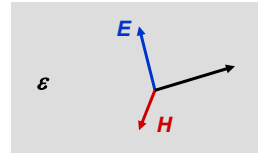


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Waves in a Dielectric Medium – Magnetic Field

Plane wave:

$$\vec{E}(\vec{r}) = \hat{n} E_0 e^{-j\vec{k}\cdot\vec{r}}$$



Calculate the magnetic field:

$$\vec{H}(\vec{r}) = \frac{j}{\omega \mu_0} \nabla \times \vec{E}(\vec{r})$$

$$\Rightarrow \vec{H}(\vec{r}) = (\hat{k} \times \hat{n}) \frac{k}{\omega \mu_0} E_0 e^{-j\vec{k}\cdot\vec{r}}$$

$$\Rightarrow \vec{H}(\vec{r}) = (\hat{k} \times \hat{n}) \frac{k}{\omega \mu_0} E_0 e^{-j\vec{k}\cdot\vec{r}}$$

$$\Rightarrow \vec{H}(\vec{r}) = (\hat{k} \times \hat{n}) \frac{E_0}{\eta} e^{-j\vec{k}\cdot\vec{r}}$$

The wave impedance also changes

$$\eta = \sqrt{\frac{\mu_0}{\epsilon}}$$

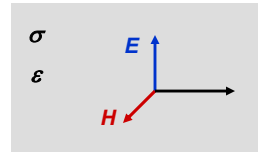
$$= \sqrt{\frac{\mu_0 \epsilon_0}{\epsilon_0 \epsilon}} = \frac{\eta_0}{n}$$

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Waves in a Conductive Medium – Complex Permittivity

Suppose we have a plane wave of the form,

$$\vec{E}(\vec{r}) = \hat{x} E_0 e^{-jkz}$$



traveling in an infinite medium with conductivity σ and permittivity ϵ

Faraday's Law:

$$\nabla \times \vec{E}(\vec{r}) = -j\omega \mu_0 \vec{H}(\vec{r})$$

Ampere's Law: Now $\vec{J}(\vec{r}) \neq 0$

$$\begin{aligned} \nabla \times \vec{H}(\vec{r}) &= \vec{J}(\vec{r}) + j\omega \epsilon \vec{E}(\vec{r}) \\ &= \sigma \vec{E}(\vec{r}) + j\omega \epsilon \vec{E}(\vec{r}) \\ &= j\omega \epsilon_{\text{eff}}(\omega) \vec{E}(\vec{r}) \end{aligned}$$

$$\epsilon_{\text{eff}}(\omega) = \epsilon \left(1 - j \frac{\sigma}{\omega \epsilon} \right)$$

Complex Wave Equation:

$$\Rightarrow \nabla^2 \vec{E}(\vec{r}) = -\omega^2 \mu_0 \epsilon_{\text{eff}}(\omega) \vec{E}(\vec{r})$$

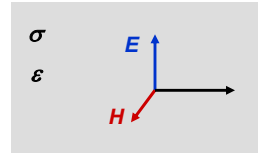
The effect of conductivity has been absorbed in a **complex frequency dependent effective permittivity**

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Waves in a Conductive Medium – Complex Refractive Index

Plane wave:

$$\vec{E}(\vec{r}) = \hat{x} E_0 e^{-jkz}$$



Dispersion relation:

$$k = \omega \sqrt{\mu_0 \epsilon_{\text{eff}}(\omega)}$$

$$\Rightarrow k = \omega \sqrt{\mu_0 \epsilon_0} \sqrt{\frac{\epsilon_{\text{eff}}(\omega)}{\epsilon_0}} \longrightarrow \left\{ \begin{array}{l} n_{\text{eff}}(\omega) = \sqrt{\frac{\epsilon_{\text{eff}}(\omega)}{\epsilon_0}} \text{ complex refractive index} \end{array} \right.$$

$$\Rightarrow k = \omega \frac{n_{\text{eff}}(\omega)}{c} \longrightarrow \left\{ \begin{array}{l} \text{compare with } k = \frac{\omega}{c} \text{ for waves in free space} \end{array} \right.$$

Since the refractive index is complex: $n_{\text{eff}}(\omega) = \sqrt{\frac{\epsilon}{\epsilon_0}} \sqrt{1 - j \frac{\sigma}{\omega \epsilon}}$
the wavevector k is also complex

$$\text{Let: } k = k' - jk''$$

↑
Real part

↑
Imaginary part

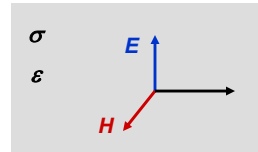
$$\frac{\sigma}{\omega \epsilon} = \text{loss tangent}$$

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Waves in a Conductive Medium – Complex Wavevector

Plane wave:

$$\vec{E}(\vec{r}) = \hat{x} E_0 e^{-jkz}$$



Complex wavevector: $k = k' - jk''$

What are the implications of a complex wavevector?

- Wave decays exponentially with distance as it propagates

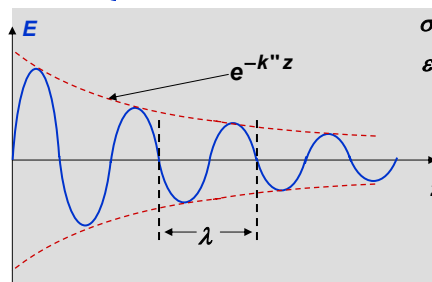
$$\vec{E}(\vec{r}) = \hat{x} E_0 e^{-jkz}$$

$$\Rightarrow \vec{E}(\vec{r}) = \hat{x} E_0 e^{-jk'z} e^{-k''z}$$

↑
exponential decay

- The wavelength is related to the real part of the wavevector:

$$k' = \frac{2\pi}{\lambda}$$



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Waves in a Conductive Medium – Magnetic Field

Plane wave:

$$\vec{E}(\vec{r}) = \hat{x} E_0 e^{-jkz}$$

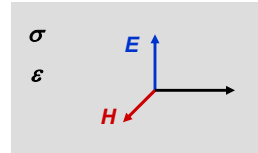
Calculate the magnetic field:

$$\vec{H}(\vec{r}) = \frac{j}{\omega \mu_0} \nabla \times \vec{E}(\vec{r})$$

$$\Rightarrow \vec{H}(\vec{r}) = \hat{y} \frac{k}{\omega \mu_0} E_0 e^{-jkz} \longrightarrow \left\{ \begin{array}{l} \text{Don't forget that the wavevector} \\ \text{"k" is complex now} \end{array} \right.$$

$$\Rightarrow \vec{H}(\vec{r}) = \hat{y} \frac{k}{\omega \mu_0} E_0 e^{-jkz}$$

$$\Rightarrow \vec{H}(\vec{r}) = \hat{y} \frac{E_0}{\eta_{\text{eff}}(\omega)} e^{-jkz} \longrightarrow \left\{ \begin{array}{l} \text{The impedance is now also} \\ \text{complex} \\ \eta_{\text{eff}}(\omega) = \sqrt{\frac{\mu_0}{\epsilon_{\text{eff}}(\omega)}} \end{array} \right.$$



Note: The E-field and the H-field are no longer in phase since $\eta_{\text{eff}}(\omega)$ is complex

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Waves in a Conductive Medium – Power Flow

Plane wave:

$$\vec{E}(\vec{r}) = \hat{x} E_0 e^{-jkz} \quad \vec{H}(\vec{r}) = \hat{y} \frac{E_0}{\eta_{\text{eff}}(\omega)} e^{-jkz}$$

$$k = k' - jk''$$

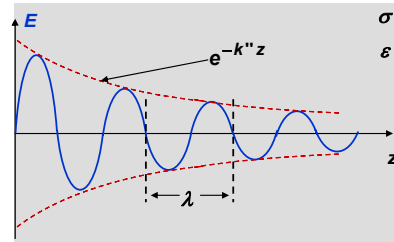
Note that: $\frac{1}{\eta_{\text{eff}}(\omega)} = \frac{k}{\omega \mu_0} = \frac{k' - jk''}{\omega \mu_0}$

Poynting vector and time average power per unit area:

$$\langle \vec{S}(\vec{r}, t) \rangle = \frac{1}{2} \text{Re}[\vec{S}(\vec{r})]$$

$$= \frac{1}{2} \text{Re}[\vec{E}(\vec{r}) \times \vec{H}^*(\vec{r})]$$

$$= \frac{1}{2} \text{Re} \left[\hat{z} \frac{E_0^2}{(\eta_{\text{eff}}(\omega))^*} e^{-2k''z} \right] = \hat{z} \frac{E_0^2}{2} \left(\frac{k'}{\omega \mu_0} \right) e^{-2k''z}$$



Time average power per unit area decays exponentially with distance because energy is dissipated in a conductive medium due to I^2R (or $J \cdot E$) type of losses and this energy dissipated is taken away from the plane wave

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Loss Tangent and Dielectric Relaxation Time - I

The complex wavevector is: $k = \omega \frac{n_{eff}(\omega)}{c}$

The complex refractive index is: $n_{eff}(\omega) = \sqrt{\frac{\epsilon}{\epsilon_0}} \sqrt{1 - j \frac{\sigma}{\omega \epsilon}}$

Loss tangent

$$\text{Loss tangent} = \frac{\sigma}{\omega \epsilon}$$

But the dielectric relaxation time was: $\tau_d = \frac{\epsilon}{\sigma}$

\Rightarrow Loss tangent = $\frac{1}{\omega \tau_d}$ And: $n_{eff}(\omega) = \sqrt{\frac{\epsilon}{\epsilon_0}} \sqrt{1 - j \frac{1}{\omega \tau_d}}$

Loss tangent

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Loss Tangent and Dielectric Relaxation Time - II

There are two possible scenarios:

High frequency and/or low conductivity case (e.g. lossy dielectrics)

$$\omega \tau_d \gg 1 \quad \text{or} \quad \frac{\omega \epsilon}{\sigma} \gg 1$$

The frequency is much greater than the inverse dielectric relaxation time

\Rightarrow The conductive medium does not have enough time to react to the electromagnetic wave

\Rightarrow No appreciable currents flow in the conductive medium

Low frequency and/or high conductivity case (e.g. Imperfect metals)

$$\omega \tau_d \ll 1 \quad \text{or} \quad \frac{\omega \epsilon}{\sigma} \ll 1$$

The frequency is much smaller than the inverse dielectric relaxation time

\Rightarrow The conductive medium has enough time to react to the electromagnetic wave

\Rightarrow Appreciable currents flow in the conductive medium

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Waves in a Conductive Medium – Lossy Dielectrics

Plane waves:

$$\vec{E}(\vec{r}) = \hat{x} E_0 e^{-jkz}$$

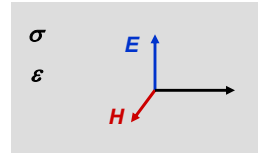
Dispersion relation

$$k = \omega \frac{n_{\text{eff}}(\omega)}{c}$$

Refractive index

$$n_{\text{eff}}(\omega) = \sqrt{\frac{\epsilon}{\epsilon_0} \left(1 - j \frac{\sigma}{\omega \epsilon} \right)}$$

$$\frac{\sigma}{\omega \epsilon} = \frac{1}{\omega \tau_d} = \text{loss tangent}$$



Lossy dielectric approximation:

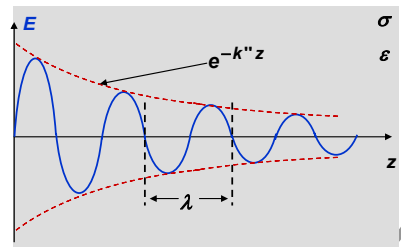
If $\frac{\sigma}{\omega \epsilon} \ll 1$ then: $n_{\text{eff}}(\omega) \approx \sqrt{\frac{\epsilon}{\epsilon_0} \left(1 - j \frac{\sigma}{2\omega \epsilon} \right)}$

$$\Rightarrow k = k' - j k'' = \frac{\omega}{c} \sqrt{\frac{\epsilon}{\epsilon_0} \left(1 - j \frac{\sigma}{2\omega \epsilon} \right)}$$

$$k' = \frac{\omega}{c} \sqrt{\frac{\epsilon}{\epsilon_0}} \quad k'' = \frac{\sigma}{2} \sqrt{\frac{\mu_0}{\epsilon}} \quad \lambda = \frac{2\pi}{k'}$$

$$\vec{E}(\vec{r}) = \hat{x} E_0 e^{-jkz}$$

$$\Rightarrow \vec{E}(\vec{r}) = \hat{x} E_0 e^{-jk'z} e^{-k''z}$$



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Waves in a Conductive Medium – Imperfect Metals

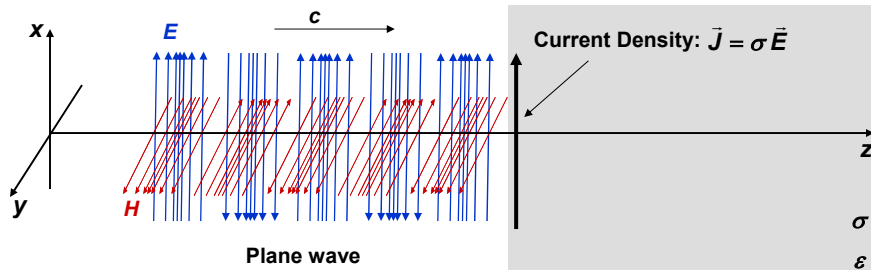
Now consider the case when:

$$\omega \tau_d \ll 1 \quad \text{or} \quad \frac{\omega \epsilon}{\sigma} \ll 1$$

The frequency is much smaller than the inverse dielectric relaxation time

⇒ The conductive medium has enough time to react to the electromagnetic wave

⇒ Appreciable currents flow in the conductive medium



These currents try to screen out the magnetic field and, therefore, prevent the electromagnetic wave from going into the conductor

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Waves in a Conductive Medium – Imperfect Metals

Plane waves:

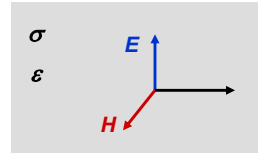
$$\vec{E}(\vec{r}) = \hat{x} E_0 e^{-jkz}$$

Dispersion relation

$$k = \omega \frac{n_{\text{eff}}(\omega)}{c}$$

Refractive index

$$n_{\text{eff}}(\omega) = \sqrt{\frac{\epsilon}{\epsilon_0} \sqrt{1 - j \frac{\sigma}{\omega \epsilon}}}$$



$$\frac{\sigma}{\omega \epsilon} = \frac{1}{\omega \tau_d} = \text{loss tangent}$$

Imperfect metal approximation:

Suppose $\frac{\sigma}{\omega \epsilon} \gg 1$ then: $n_{\text{eff}}(\omega) \approx \sqrt{\frac{\epsilon}{\epsilon_0} \sqrt{-j \frac{\sigma}{\omega \epsilon}}} = \sqrt{\frac{\sigma}{2 \omega \epsilon_0}} (1 - j)$

$$\Rightarrow k = k' - j k'' = \sqrt{\frac{\sigma \omega \mu_0}{2}} (1 - j) = \frac{1}{\delta} (1 - j)$$

$$k' = \sqrt{\frac{\omega \mu_0 \sigma}{2}} = \frac{1}{\delta} \quad k'' = \sqrt{\frac{\omega \mu_0 \sigma}{2}} = \frac{1}{\delta} \quad \lambda = \frac{2\pi}{k'} = 2\pi \delta$$

$\delta = \text{penetration depth or skin-depth}$

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Waves in a Conductive Medium – Imperfect Metals

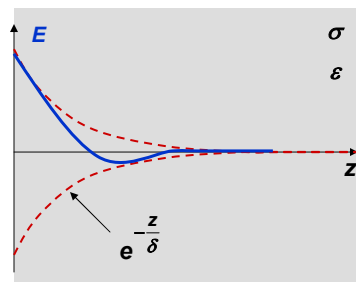
Due to current screening the wave decays within a few skin-depths:

$$\vec{E}(\vec{r}) = \hat{x} E_0 e^{-jkz}$$

$$\Rightarrow \vec{E}(\vec{r}) = \hat{x} E_0 e^{-jk'z} e^{-k''z}$$

$$\Rightarrow \vec{E}(\vec{r}) = \hat{x} E_0 e^{-jk'z} e^{-\frac{z}{\delta}}$$

$\delta = \text{penetration depth or skin-depth}$



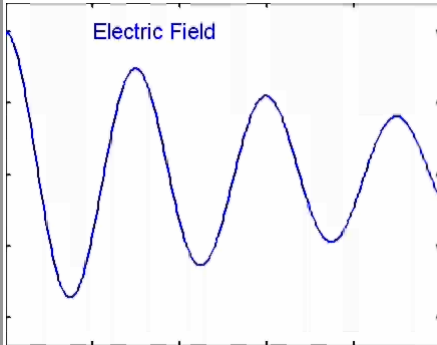
Since the wavelength λ inside the medium is $2\pi\delta$, the wave hardly propagates one wavelength distance into the medium

The screening current density, given by $\vec{J} = \sigma \vec{E}$, is non-zero only in a layer of thickness equal to skin-depth δ near the surface

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Waves in a Conductive Medium

Lossy Dielectrics



Imperfect Metals

